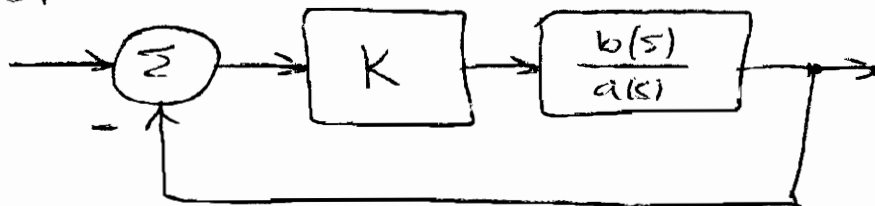


## Root Locus Rules

Consider



Note: e.g. if  $C(s) = \frac{K(s+1)}{s+2}$ ,  $G(s) = \frac{5}{s+10}$

$$\text{then } \frac{b(s)}{a(s)} = \frac{5(s+1)}{(s+2)(s+10)}$$

In general,  $\frac{b(s)}{a(s)}$  has  $m$  zeros and  $n$  poles

We are interested in the poles of the CLTF as  $K$  varies from 0 to  $\infty$ . The plot of the pole locations is called a root locus.

Rule 0 The root locus is symmetric about the real axis

Rule 1 The  $n$  branches of the locus start at the open-loop poles (roots of  $a(s)$ ) and move toward the  $m$  open-loop zeros (roots of  $b(s)$ ) with the remaining  $n-m$  branches going to infinity.

Rule 2 The branches on the real axis lie to the left of an odd number of open loop poles and zeros

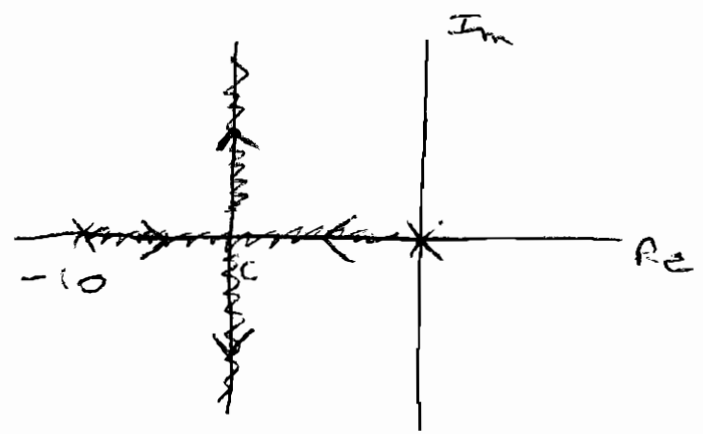
Rule 3  $n-m$  branches go to infinity at asymptotes centered at

$p_i = \text{roots of } a(s)$        $c = \frac{\sum p_i - \sum z_i}{n-m}$   
 $z_i = \text{roots of } b(s)$

at angles  $\phi_l = \frac{180^\circ + (l-1)360^\circ}{n-m}$ ,  $l=1, 2, \dots, n-m$

Example 1:  $\frac{b(s)}{a(s)} = \frac{5}{s(s+10)}$

$m=0$ , no  $z_i$   
 $n=2$ ,  $p_1=0$   
 $p_2=-10$



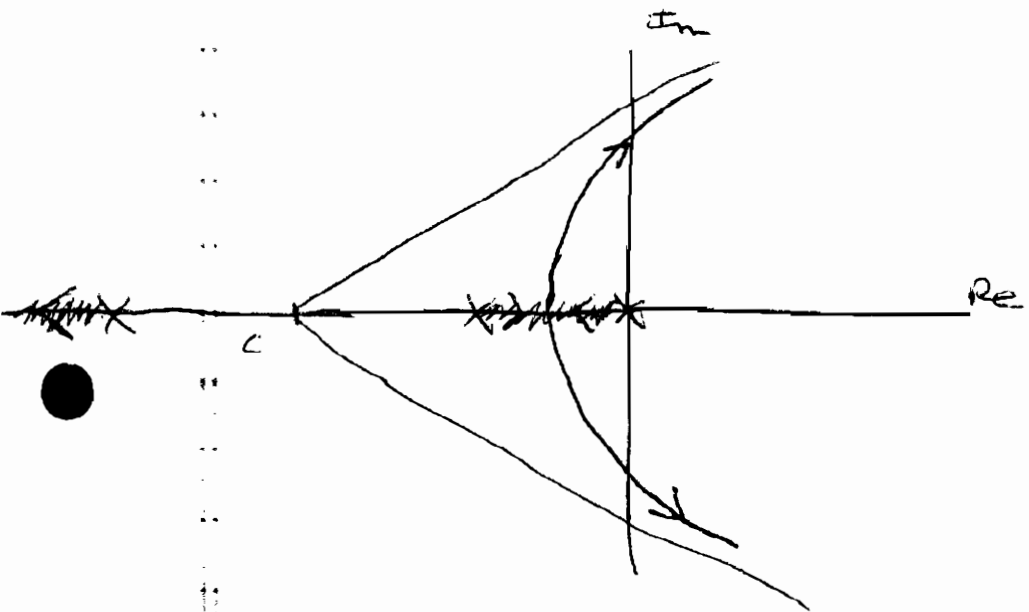
$c = \frac{-10}{2} = -5$

$\phi_1 = 90^\circ$

$\phi_2 = 270^\circ$

Example 2:  $\frac{b(s)}{a(s)} = \frac{5}{s(s+10)(\frac{s}{100}+1)}$

$m=0$   
 $n=3$   
 $p_1=0$   
 $p_2=-10$   
 $p_3=-100$



$c = \frac{-110}{3} \approx -37$

$\phi_1 = 60^\circ$

$\phi_2 = 180^\circ$

$\phi_3 = 300^\circ$

Example 3  $\frac{b(s)}{a(s)} = \frac{s+5}{s^2}$

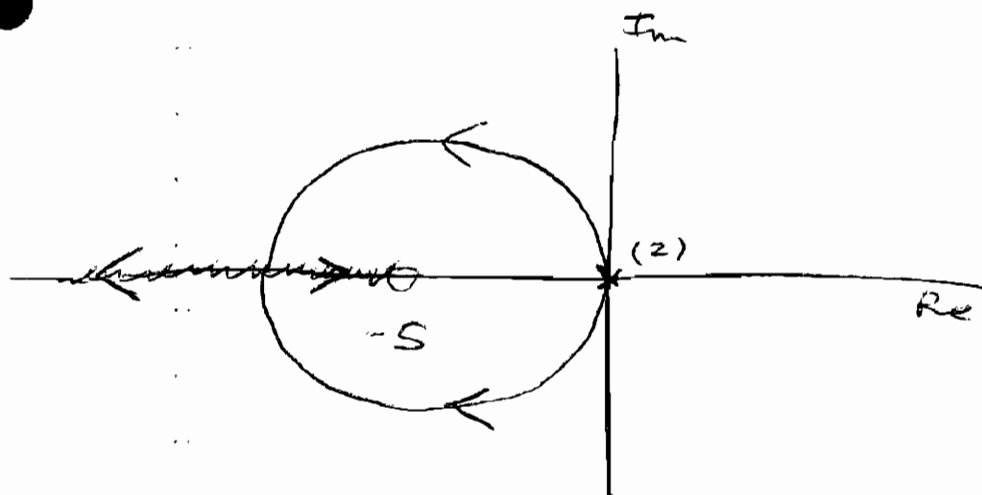
$$m=1 \quad z_1 = -5$$

$$p=2 \quad p_1 = 0$$

$$p_2 = 0$$

$$c = 5$$

$$\phi_1 = 180^\circ$$



Example 4  $\frac{b(s)}{a(s)} = \frac{s+5}{s^2(s+10)}$

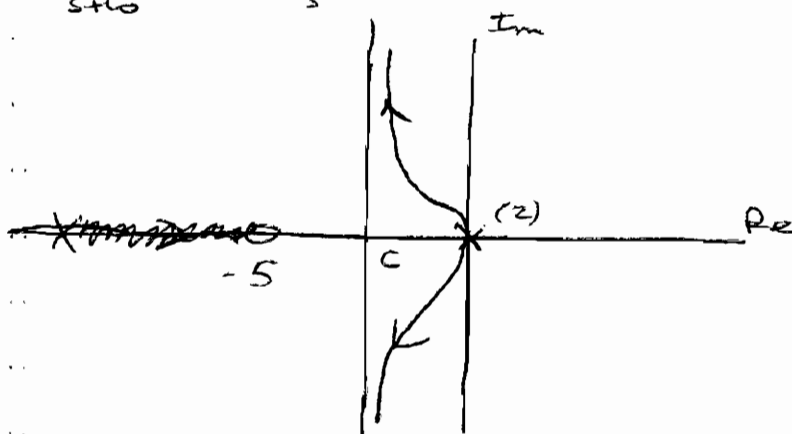
$$C(s) = k \frac{s+5}{s+10}, \quad G(s) = \frac{1}{s^2}$$

$$m=1, \quad n=3$$

$$c = -\frac{5}{2} = -2.5$$

$$\phi_1 = 90^\circ$$

$$\phi_2 = 270^\circ$$



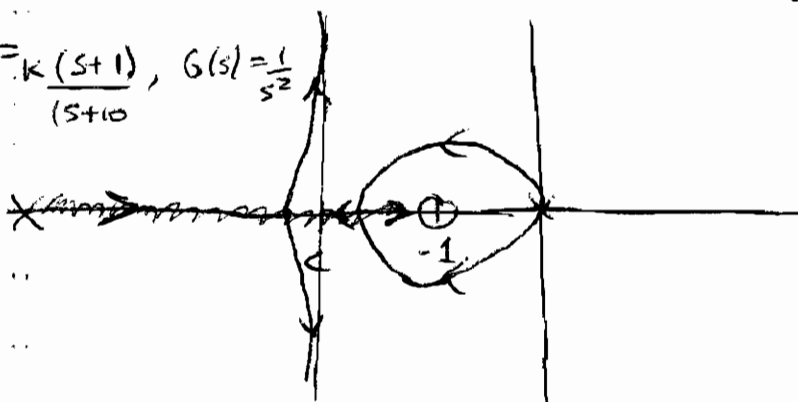
Example 5  $\frac{b(s)}{a(s)} = \frac{s+1}{s^2(s+10)}$

$$C(s) = k \frac{(s+1)}{(s+10)}, \quad G(s) = \frac{1}{s^2}$$

$$m=1, \quad n=3$$

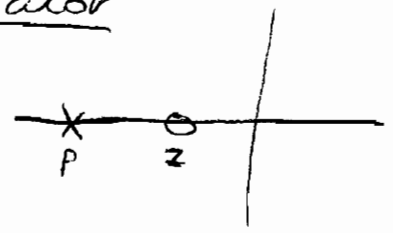
$$c = -\frac{9}{2} = -4.5$$

$$\phi_{1,2} = \frac{90^\circ}{270^\circ}$$



## Phase-Lead Compensator

$$C(s) = K \frac{(s-z)}{(s-p)}$$

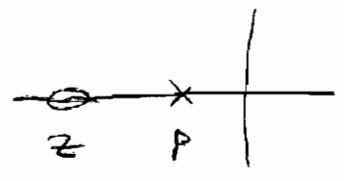


Effect: pulls the root locus  
 [settling time] "to the left" resulting  
 in a stable, "fast" system.

Design Procedure: Get a root locus plot of desired shape. Use Matlab to meet numerical specifications.  
 Consider effect of  $p \neq z$  on center of asymptotes

## Phase Lag Compensator

$$C(s) = K \frac{(s-z)}{(s-p)}$$

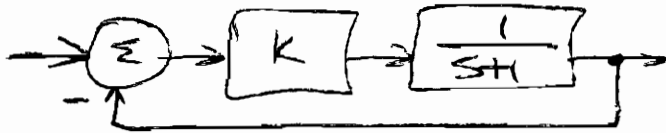
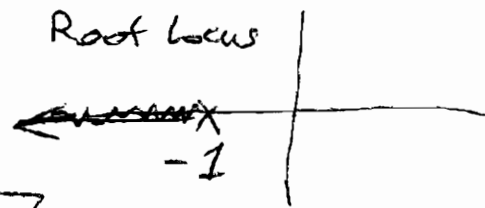


Effect: pulls root locus "to the right".  
 [steady-state error] Thus, need to be careful about stability.  
 This is why phase lag is needed →  $\times$  May be necessary to reduce steady-state error.

If we let  $p=0$  then  $C(s)$  contains a pure integrator, and this special phase-lag compensator is called a PI (proportional plus integral) compensator.  
 A PI compensator increases the system type by 1.

Example:  $C(s) = K$      $G(s) = \frac{1}{s+1}$

$$\frac{b(s)}{a(s)} = \frac{1}{s+1}$$



Suppose we want  $T_s = 2$  sec

CLTF is  $\frac{K}{s+1+K}$  . Pole @  $s = -(1+K)$

$$T_s = 4.62/\alpha \Rightarrow \alpha = 2.31 \Rightarrow \text{pole} = -2.31$$

$$1+K = 2.31$$

$$K = 1.31$$

For this value of  $K$  what is the steady-state error to a step of height  $R$ ?

$$Y_{ss} = (\text{DC gain}) \cdot (R) = \frac{1.31}{1+1.31} \cdot R = .57R$$

$$e_{ss} = R - Y_{ss} = (1 - .57)R = .43R$$

the steady-state error is 43% of the reference input.

Suppose we want  $e_{ss} = 0$  and  $T_s = 2$ ?  
To get  $e_{ss} = 0$  we need a Type 1 system.

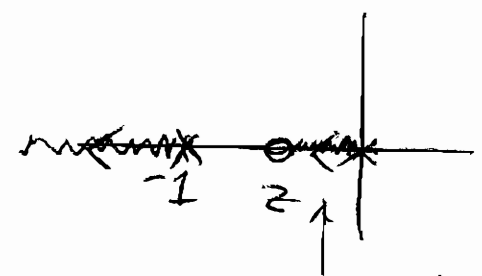
Let  $C(s) = K \cdot \frac{(s-z)}{s}$  i.e. a phase lag compensator  
PI Compensator (pole to right of zero)

$$\frac{b(s)}{a(s)} = \frac{s-z}{s(s+1)}$$

Two possibilities



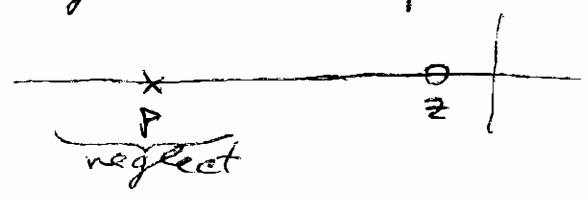
This will work well for  $z \approx -2$



This pole is too slow

In Matlab, check  $\frac{b(s)}{a(s)} = \frac{s+2}{s(s+1)}$

A phase-lead compensator  $K \frac{(s-z)}{(s-p)}$  with a large value of  $p$



at low frequencies  $\approx k'(s-z) = \underset{\substack{\uparrow \\ \text{derivative}}}{k's} - \underset{\substack{\uparrow \\ \text{proportional}}}{kz}$

This is a PD (proportional plus derivative) compensator.

A practical PID (proportional/integral/derivative) compensator is a combination of phase lag (PI) and phase lead:

$$\text{PID: } C(s) = \frac{k(s-z_1)(s-z_2)}{s(s-p)}$$