Z-Transform and Digital Systems

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Digital systems

Figure on the right shows a typical discrete-time system for processing real-world (analog) signals. The analog input x(t) is digitized by the analog-to-digital converter (A/D). The resulting digital input x[n] is sent to a digital linear time-invariant system (LTI) for processing. The output digital signal y[n] is sent to the digital-to-analog converter (D/A), which generates the analog output y(t).



The sampling rate f_s set forth by the A/D is arguably the most crucial parameter in this digital system. Sampling is almost always done periodically with a sampling period of *T*.

$$f_s = \frac{1}{T}$$

The digital (or discrete-time) signal is a sequence of numbers. The actual time axis should be denoted as ... -3T, -2T, -T, 0, *T*, 2*T*, 3*T*, ..., where *T* is the sampling period. Without losing generality we drop *T* and just use the index *n* to represent the time axis. As shown in the figure, The impulse is the delta function that takes the value of 1 at *n* = 0, and 0 elsewhere. The output in response to the Kronecker $\delta[n]$ is the impulse response h[n]. The impulse response h[n] completely characterizes a discrete-time LTI system, because the output is the convolution of the input and the impulse response:

$$y[n] = x[n] \otimes h[n]$$

$$y[n] = \sum_{m=-\infty}^{\infty} x[m]h[n-m] = \sum_{m=-\infty}^{\infty} x[n-m]h[m]$$

The digital equivalence to the Laplace transform (LT) is the z-transform. The z-transform (ZT) and the inverse z-transform (IZT) are defined as follows:

ZT:
$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

IZT:
$$x[n] = \frac{1}{2\pi j} \oint X(z) z^{n-1} dz$$

where $z = r e^{j\omega}$, $\left| \sum_{-\infty}^{\infty} x[n] z^{-n} \right| < \infty$

The relationship between LT and ZT is illustrated in the figure. The $j\omega$ axis on the s-plane is mapped onto the unit circle $e^{j\omega}$ on the z-plane. The left-hand-side of the s-plane is mapped onto the inside of the unit circle.



 $2 - \frac{\delta[n]}{1 - \frac{1}{2} - \frac{1}{2}$

The Fourier transform is the ZT evaluated on the unit circle. The Fourier transform (FT) and the inverse Fourier transform (IFT) are defined as follows:

FT:
$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

IFT: $x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})e^{j\omega n} d\omega$

Duality

Duality refers to certain properties of the linear transforms that exhibit symmetry between the time domain and the frequency domain.

	Time domain	Frequency domain
Duality	Time-limited	Band-unlimited
	Time-unlimited	Band-limited
Duality	Periodic	Discrete
	Discrete	Periodic

A periodic signal consists of its fundamental and harmonics, but no other frequencies. Therefore, its frequency spectrum is discrete. Based on duality, a discrete signal has a periodic frequency spectrum. One way to understand this is that the discrete Fourier transform is evaluated around the unit circle. Thus, the frequency spectrum of a discrete signal has a period of 2π . This is also related to the Nyquist sampling theory, discussed in more detail below.

Sampling theory (aka Nyquist sampling theory, Nyquist-Shannon sampling theory)

A sufficient sampling frequency to represent a signal is at least higher than twice the highest frequency component of the signal. To illustrate this, consider the following example: A sine wave has frequency f_0 . If we choose a sampling rate right at the Nyquist limit, which is $2f_0$, we get two sample points on each cycle. We could get all 0's and fail to represent the signal. However, if we sample at a rate > f_0 , we will avoid the straight line scenario and can represent that frequency.

Sampling rate $f_s = 2 f_0$ $\sin 2\pi f_0 t$ Sampling rate $f_s > 2 f_0$