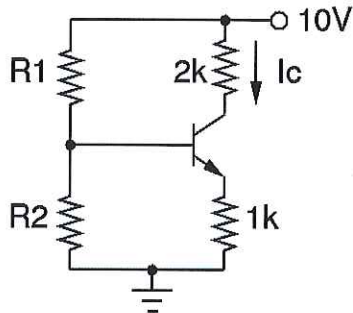


Name .....

Score .....

1) BJT Biasing



**BJT Parameters:**

$$V_T = 30 \text{ mV}$$

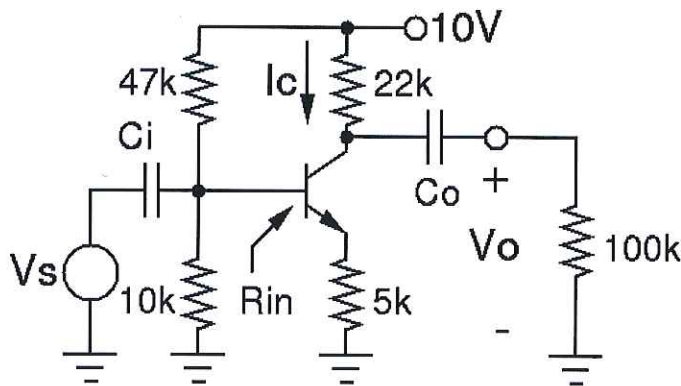
$$V_{BEQ} = 0.7 \text{ V}$$

$$V_A = \infty$$

$$\beta = 100$$

- If  $R_2 = 2.7 \text{ k}\Omega$ , find a value for  $R_1$  such that  $I_{CQ} = 2 \text{ mA}$ .
- If the equivalent base resistor  $R_B$  is  $4 \text{ k}\Omega$  (not the solution of the above question), how much would  $I_{CQ}$  change (in %) if you were to double the values of  $R_1$  and  $R_2$ ?
- If  $R_B$  is  $4 \text{ k}\Omega$ , how much would  $I_{CQ}$  change (in %) if you were to double the value of  $R_E$ ?
- If  $R_2 = \infty$  while  $R_E$  is kept at  $1 \text{ k}\Omega$ , find a value for  $R_1$  such that  $I_{CQ}$  remains at  $2 \text{ mA}$ .

2) Common Emitter Amplifier



**BJT Parameters:**

$$V_T = 30 \text{ mV}$$

$$V_{BEQ} = 0.7 \text{ V}$$

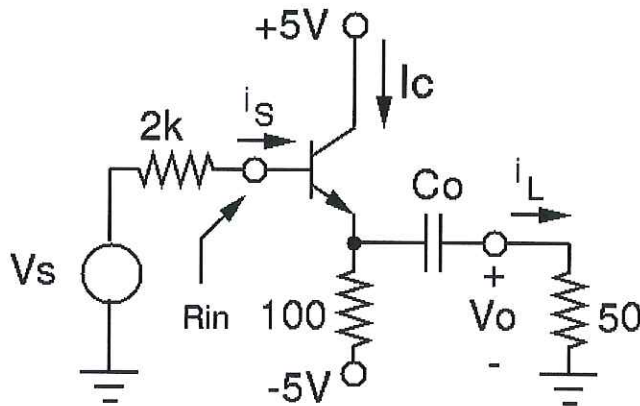
$$V_A = \infty$$

$$\beta = 100$$

Assume that the 2 caps act as AC shorts for the frequency range of interest.

- Derive values for  $I_{CQ}$  and  $V_{CEQ}$ .
- Sketch the small signal equivalent circuit and find a value for the transistor input resistor  $r_{ie}$  or  $r_{\pi e}$  (Describe the transistor collector current as  $\beta i_b$  and clearly mark  $i_b$ ).
- Derive a value for the input resistance  $R_{in}$  seen at the base of the transistor.
- Derive a value for the voltage gain  $A_V = v_o/v_s$ .

### 3) Common Collector Amplifier



#### BJT Parameters:

$$V_T = 30 \text{ mV}$$

$$V_{BEQ} = 0.7 \text{ V}$$

$$V_A = \infty$$

$$\beta = 100$$

$V_s$  is an **ideal AC voltage source**.

- Find **values** for the transistor Q-point parameters  $I_{CQ}$  and  $V_{CEQ}$ .
- Assuming linear operation, what is the **maximum amplitude of  $v_o$**  this amplifier can support?
- Derive an **expression** and a **value** for the **input resistance  $R_{in}$**  seen at the base of the transistor (Assume that  $C_o$  acts as a short for the frequency range of interest).
- Find a **value** for the **AC current gain  $A_I = i_L / i_s$**  of the depicted amplifier. (Assume that  $C_o$  acts as a short for the frequency range of interest).

$$1) a) \left| I_{CQ} \approx \frac{V_{CC} \frac{\pi_2}{\pi_1 + \pi_2} - V_{BEQ}}{\frac{\pi_1 \pi_2}{\beta(\pi_1 + \pi_2)} + \pi_E} = \frac{27}{\pi_1 + 2.7} - 0.7 \text{ [mA]} \right|$$

$$\therefore 27 + 0.7(\pi_1 + 2.7) = 0.054\pi_1 + 2\pi_1 + 5.4$$

$$\left| \pi_1 = \frac{19.71}{2.754} \text{ [k}\Omega\text{]} = 7.16 \text{ k}\Omega \right| \quad 3$$

$$b) \pi_{B1} = 4 \text{ k}\Omega \quad \pi_{B2} = 8 \text{ k}\Omega$$

$$\therefore \left| \frac{I_{CQ2}}{I_{CQ1}} = \frac{\frac{\pi_{B1}}{\beta} + \pi_E}{\frac{\pi_{B2}}{\beta} + \pi_E} = \frac{40 + 1,000}{80 + 1,000} \approx 0.96 \right|$$

$$\therefore \left| I_{CQ} \text{ would decrease by } 4\% \right| \quad 3$$

$$c) \pi_{E1} = 1 \text{ k}\Omega \quad \pi_{E2} = 2 \text{ k}\Omega$$

$$\left| \frac{I_{CQ2}}{I_{CQ1}} = \frac{\frac{\pi_B}{\beta} + \pi_{E1}}{\frac{\pi_B}{\beta} + \pi_{E2}} = \frac{40 + 1,000}{40 + 2,000} \approx 0.51 \right|$$

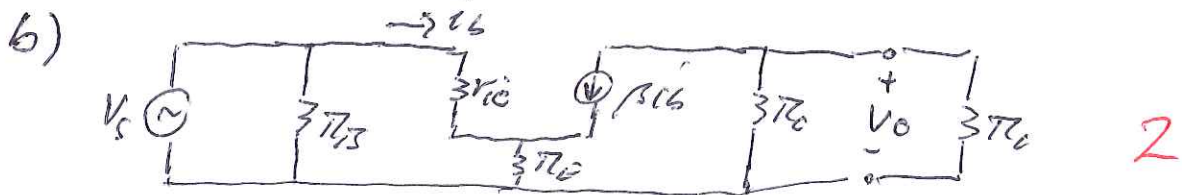
$$\therefore \left| I_{CQ} \text{ would decrease by } 49\% \right| \quad 3$$

$$d) \left| I_{CQ} = \frac{V_{CC} - V_{BEQ}}{\frac{\pi_1}{\beta} + \pi_E} = \frac{10 - 0.7}{\frac{\pi_1}{100} + 1} \text{ [mA]} \right|$$

$$\therefore \left| \pi_1 = 100 \left[ \frac{9.3}{2} - 1 \right] \text{ [k}\Omega\text{]} = 365 \text{ k}\Omega \right| \quad 3$$

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$$2) a) \left\| \begin{aligned} I_{CQ} &\approx \frac{V_{CC} \frac{\pi_C}{\pi_C + \pi_E} - V_{BEQ}}{\frac{\pi_C \pi_E}{\beta(\pi_C + \pi_E)} + \pi_E} \approx 0.21 \text{ mA} \\ V_{CEQ} &\approx V_{CC} - I_{CQ}(\pi_C + \pi_E) \approx 4.3 \text{ V} \end{aligned} \right\| \begin{array}{l} 2 \\ 1 \end{array}$$



$$\left\| r_{ie} = \frac{\beta}{g_m} = \beta \frac{V_T}{I_{CQ}} \approx 14.3 \text{ k}\Omega \right\| \quad 1$$

$$\left\| \pi_B = \pi_C \parallel \pi_E \approx 8.25 \text{ k}\Omega \right\|$$

c)  $\left\| R_{in} = r_{ie} + (1 + \beta)\pi_E \right\|$

$$\therefore \left\| \pi_{in} = 14.3 \text{ k}\Omega + 505 \text{ k}\Omega \approx 519 \text{ k}\Omega \right\| \quad 3$$

d)  $\left\| \begin{aligned} v_s &= i_b (r_{ie} + (1 + \beta)\pi_E) \\ v_o &= -\beta i_b \pi_C \parallel \pi_L \end{aligned} \right\|$

$$\therefore \left\| A_v = \frac{v_o}{v_s} = - \frac{\beta \pi_C \parallel \pi_L}{r_{ie} + (1 + \beta)\pi_E} \right\|$$

$$\therefore \left\| A_v \approx - \frac{g_m \pi_C \parallel \pi_L}{1 + g_m \pi_E} \approx -3.5 \right\| \quad 3$$

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$$3) a) \quad \left\| \begin{aligned} I_{CQ} &\approx \frac{V_{B3} - V_{BEQ}}{\frac{\beta}{\beta + 1} R_E} = \frac{5 - 0.7}{20 + 100} \text{ mA} \approx 35 \text{ } \mu\text{mA} \\ V_{CEQ} &= V_{CC} + V_{EE} - I_{CQ} R_E \approx 6.44 \text{ V} \end{aligned} \right\| \begin{array}{l} 2 \\ 1 \end{array}$$

$$b) \quad \left\| V_{\text{output}} = I_{CQ} \cdot R_E \parallel R_L \approx 1.19 \text{ V} \right\| 3$$

$$c) \quad \left\| \pi_{in} = r_{ic} + (1 + \beta) R_E \parallel R_C \right\|$$

$$\therefore \left\| \pi_{in} = \frac{r_{ic}}{\beta} + (1 + \beta) R_E \parallel R_C \approx 3.45 \text{ k}\Omega \right\| 3$$

$$d) \quad \left\| \begin{aligned} KCL: \quad i_e + i_L &= (1 + \beta) i_S \\ KVL: \quad i_e R_E &= i_L R_C \end{aligned} \right\|$$

$$\therefore \left\| i_L \left( \frac{R_C}{R_E} + 1 \right) = (1 + \beta) i_S \right\|$$

$$\therefore \left\| A_I = \frac{i_L}{i_S} = \frac{1 + \beta}{R_C/R_E + 1} \approx 67.5 \right\| 3$$

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