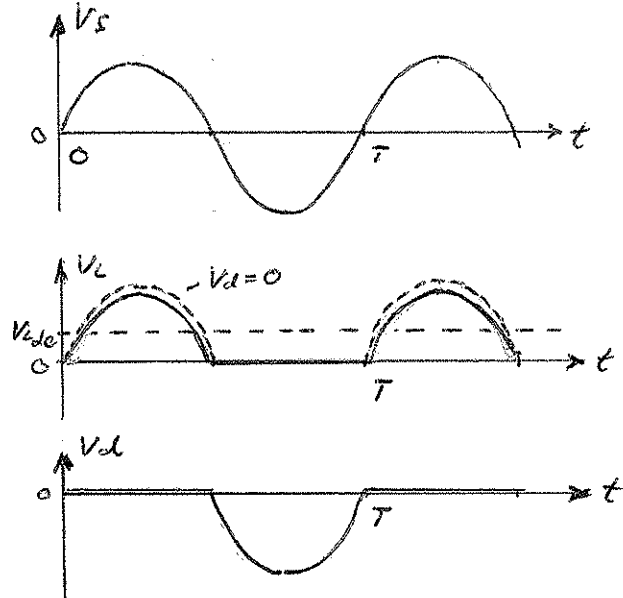
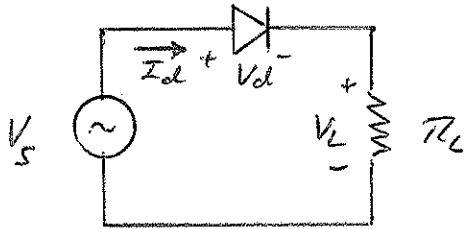


IV. Nonlinear Circuit Applications

1. Rectifiers

a) Half-wave Rectifier



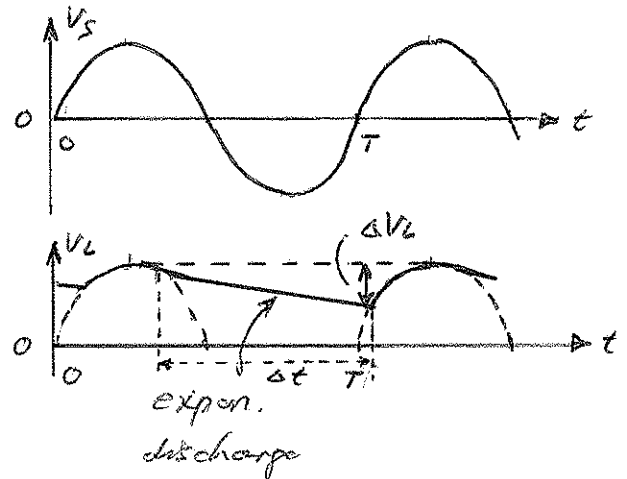
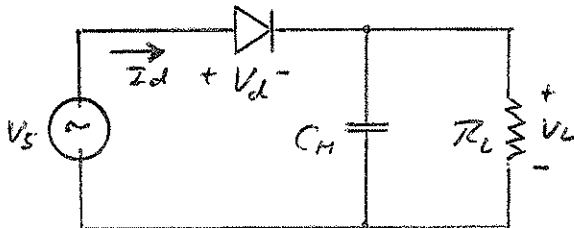
KVL:  $V_S = V_d + V_L$

$\therefore |V_L = V_S - V_d|$

dc Component of  $V_L$ :

$$\| V_{Ldc} = \frac{1}{2\pi} \int_0^{\pi} \hat{V}_L \sin(\phi) d\phi = \frac{\hat{V}_L}{2\pi} [-\cos\phi]_0^{\pi} = \frac{1}{\pi} \hat{V}_L \|$$

Half-wave Rectifier with Hold Capacitor



Ripple  $\Delta V_L$

$$|\Delta V_L = \hat{V}_S (1 - e^{-\frac{\Delta t}{\tau}}) \approx \hat{V}_S \frac{\Delta t}{\tau}|$$

where

$$\tau = \tau_L \cdot C_H$$

$$\Delta t \approx T$$

$$\therefore |\Delta V_L \approx \hat{V}_S \frac{T}{\tau_L C_H}|$$

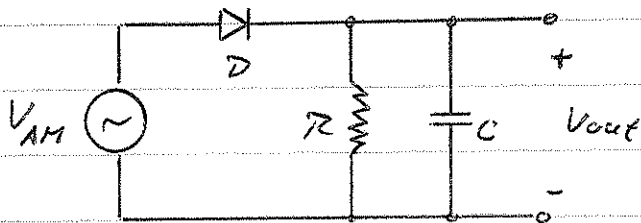
To obtain a smooth output voltage with little ripple,

$$\tau_L C_H \gg T$$

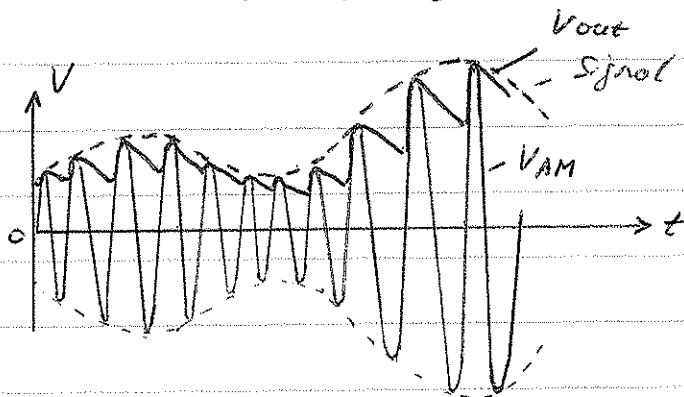
## Important application of a half-wave rectifier

AM Demodulator (Envelope detector)

### Basic Configuration



$$\tau = \tau C$$



$$|V_{AM} = A(t) \cos \omega_c t|$$

$\omega_c$ : carrier frequency

$$A(t) \approx A_0 \cos \omega_0 t$$

$\omega_0$ : message sign. freq.

The  $\tau C$  time constant  $\tau$  of the detector must be chosen such that the maximum slope of the output voltage is at least as large as the maximum slope of the envelope, i.e. the AM input signal before modulation.

## IV - 3

Thus

$$\text{Max} \left\{ \frac{d}{dt} [A_0 \cos \omega_0 t] \right\} < \text{Max} \left\{ \left| \frac{d}{dt} [A_0 e^{-\frac{t}{\tau C}}] \right| \right\}$$

$$A_0 \omega_{\text{max}} < A_0 \frac{1}{\tau C}$$

$$\therefore \left\| \tau C < \frac{1}{\omega_{\text{max}}} \right\|$$

On the other hand, the  $\tau C$  time constant should not be smaller than the period of the carrier, since then the hold effect would be very weak.

$$\therefore \left\| \tau C > \frac{2\pi}{\omega_{\text{min}}} \right\|$$

We thus obtain the following condition for the time constant  $\tau = \tau C$  of the envelope detector

$$\boxed{\frac{2\pi}{\omega_{\text{min}}} < \tau C < \frac{1}{\omega_{\text{max}}}}$$

AM Radio:  $0.5 \text{ MHz} \leq f_c \leq 1.6 \text{ MHz}$   
 $f_0 \leq 4.5 \text{ kHz}$

$$\therefore \left\| 2\mu\text{s} < \tau C < 35.4\mu\text{s} \right\|$$

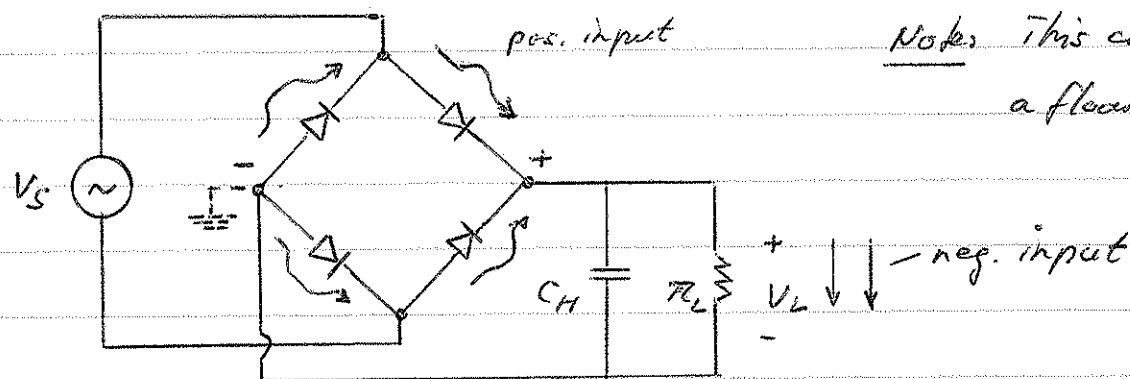
Choose  $\tau C$  as the geometric mean of the upper and lower bound

$$\left\| \tau C \approx 8.4\mu\text{s} \right\|$$

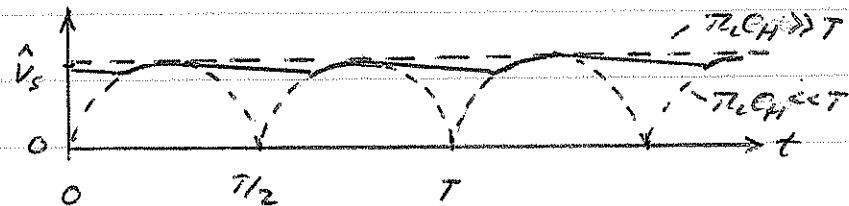
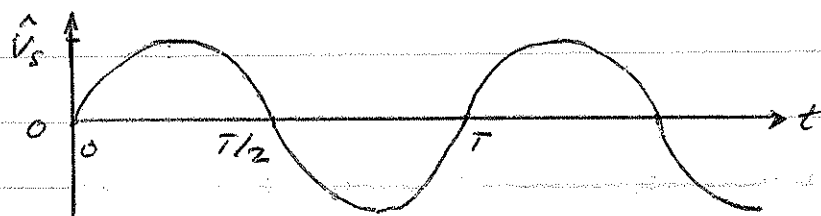
e.g.  $\left| \begin{array}{l} \tau = 1\mu\text{s} \\ C = 8.2\text{nF} \end{array} \right|$

b) Full-wave Rectifier

b1) Bridge Full-wave Rectifier



Notes This circuit requires a floating source.

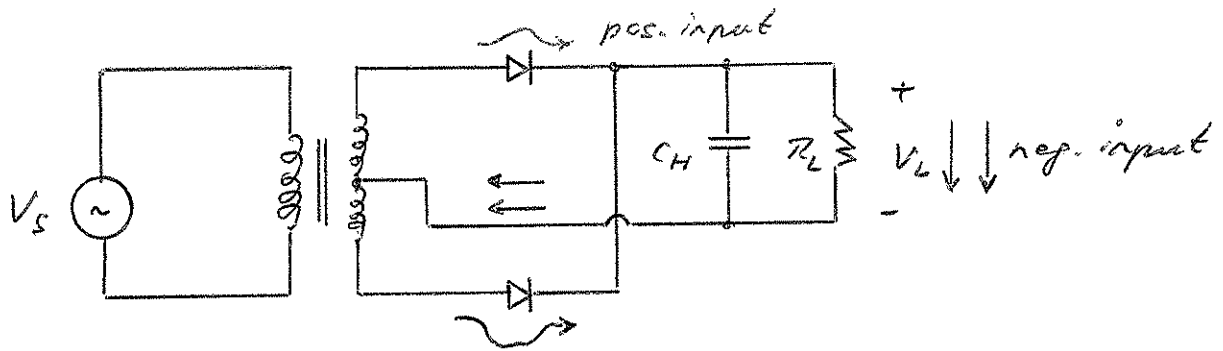


ripple is a function of the ratio  $\frac{1}{C_H T_L}$

In order to obtain a completely smooth dc output voltage, one generally applies a voltage regulator circuit to the output of the rectifier.

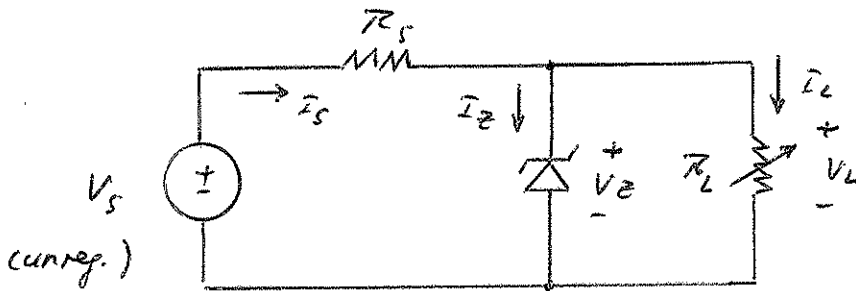
An alternative solution is to employ a lowpass filter which cuts off all harmonics of the input frequency.

62) Center tapped transformer full-wave rectifier



2. Voltage Regulators

Basic Configuration (with Zener diode)



A voltage regulator can be used to remove a ripple from an input voltage and (or) to maintain a const. output voltage over a range of loads.

Example 1       $V_s = 20V$  (const.)  
                           $V_z = 10V$

Determine  $R_s$  so that  $V_L$  remains constant at  $10V$  while the load resistor  $R_L$  varies between  $R_{Lmin} = 100\Omega$  and  $R_{Lmax} = 1k\Omega$ .

$$\underline{IV-6}$$

Solution KVL:  $V_S = I_S \cdot R_S + V_Z$

$$\therefore \left| R_S = \frac{V_S - V_Z}{I_S} \right|$$

KCL:  $I_S = I_Z + I_L = \text{const. (for } V_S = \text{const.)}$

Condition that  $V_L = V_Z \quad \left| I_Z > 0 \right|$  otherwise  $V_L < V_Z$

$$\therefore I_S > I_{L \max} = \frac{V_Z}{R_{L \min}}$$

$$\therefore R_S < \frac{V_S - V_Z}{I_{L \max}} = \frac{V_S - V_Z}{V_Z} R_{L \min}$$

$$\left\| R_S < 100 \Omega \right\|$$

Note: If  $R_L = R_{L \max}$ , we obtain a minimum load current, hence  $I_Z > 0$  is met under all conditions

if we select  $R_S = 90 \Omega$ , then

$$\left| \begin{array}{l} I_{Z \min} = 11.11 \text{ mA} \\ I_{Z \max} = 101.11 \text{ mA} \end{array} \right|$$

The Zener diode must therefore be capable of dissipating a maximum power of

$$\left| P_{Z \max} = V_Z \cdot I_{Z \max} = 1.011 \text{ W} \right|$$

$$P_{Z \max} = V_Z \cdot \left( \frac{V_S - V_Z}{R_S} - \frac{V_Z}{R_{L \max}} \right)$$

if  $P_{Z \max} < 1 \text{ W}$  then increase  $R_S$

$$R_S = 95 \Omega$$

$$I_{Z \min} = 5.3 \text{ mA}$$

$$I_{Z \max} = 95.3 \text{ mA}$$

$$\therefore \left| P_{Z \max} = 0.95 \text{ mW} \right|$$

## IV-71

example 2:  $6V \leq V_s \leq 7.5V$

$$100\Omega \leq R_L \leq \infty$$

$$V_Z = 5V$$

Determine  $R_s$  and the max power dissipated by the Zener diode.

Solution: KVL:  $V_s = I_s R_s + V_Z$

$$\text{KCL: } I_s = I_L + I_Z$$

Condition to maintain  $V_L = V_Z$

$$I_{s\text{min}} = \frac{V_{s\text{min}} - V_Z}{R_s} \geq \frac{V_Z}{R_{L\text{min}}} = I_{L\text{max}}$$

$$\therefore \left\| R_s \leq R_{L\text{min}} \frac{V_{s\text{min}} - V_Z}{V_Z} = 20\Omega \right\|$$

select  $R_s = 15\Omega$

$$\therefore \left\| I_{i\text{max}} = \frac{V_{s\text{max}} - V_Z}{R_s} \approx 0.167A \right\|$$

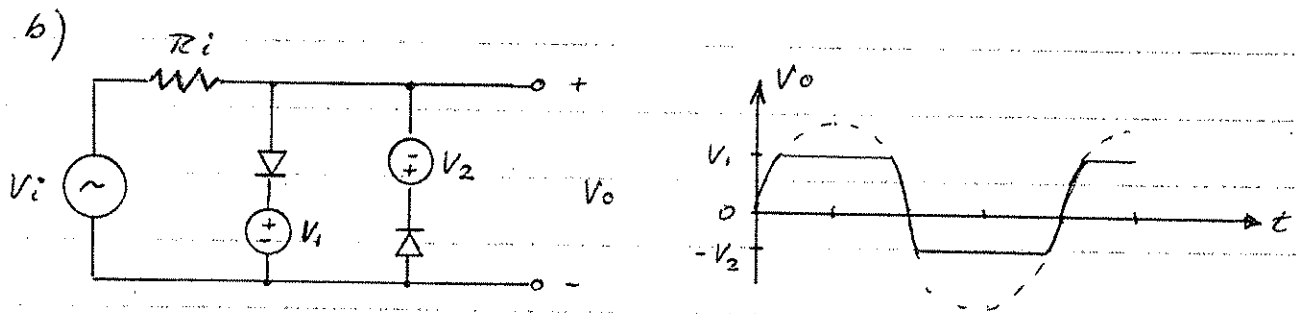
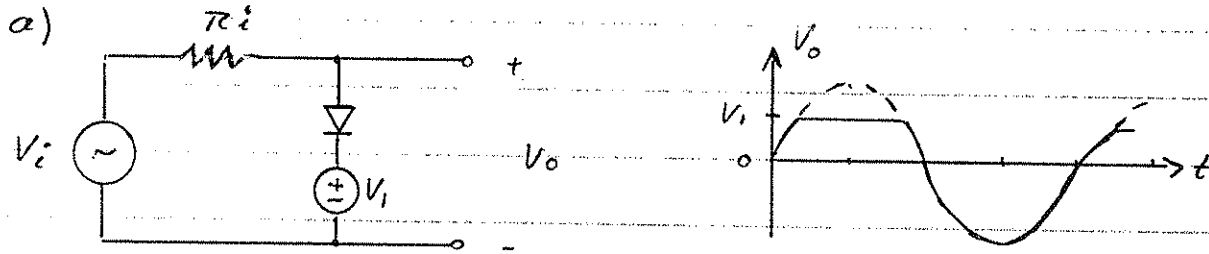
$$\left\| P_{Z\text{max}} = V_Z \left( I_{i\text{max}} - \frac{V_Z}{R_{L\text{max}}} \right) \approx 0.833W \right\|$$

$$\text{check } I_{Z\text{min}} = I_{s\text{min}} - I_{L\text{max}} = \frac{V_{s\text{min}} - V_Z}{R_s} - \frac{V_Z}{R_{L\text{min}}} = 0.017A > 0$$

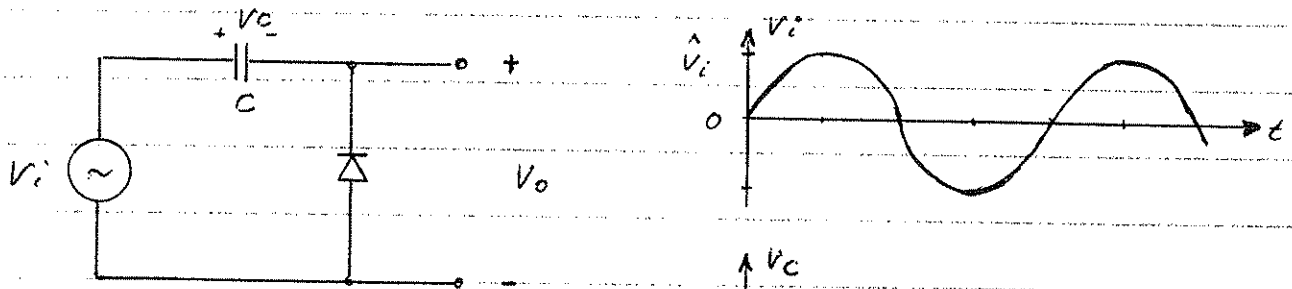
Thus  $V_L = V_Z$  under worst case condition!

### 1.3.4 Clipping and Clamping

Clipping circuits are used to limit voltage excursions.

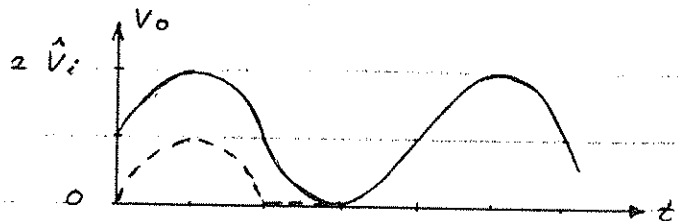


Clamping is used to make sure a voltage never goes negative (or positive.)



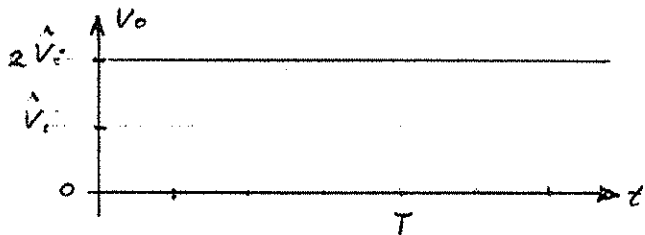
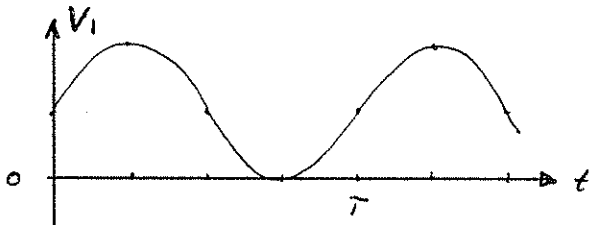
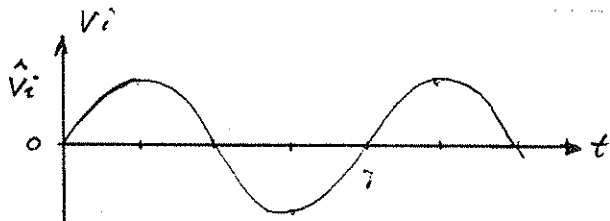
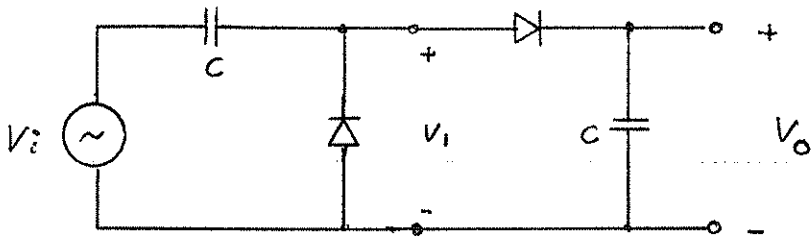
$$V_i = V_c + V_o$$

$$\Rightarrow \underline{V_o = V_i - V_c}$$





Cascade of clamping circuit with simple rectifier



voltage doubling

Voltage multiplier

