

SINCE  $x_i$  NOT DEPENDENT ON OTHER  
OUTPUTS  $\Rightarrow$  CALLED DISCRETE

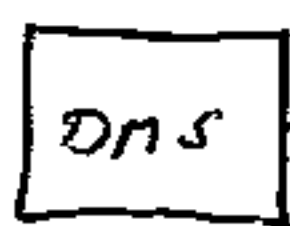
MEMORYLESS SOURCE

$\uparrow$   
INDEPENDENCE

$\uparrow$  IN TIME  
AND AMPLITUDE

OR DMS

EXAMPLE : ENGLISH ALPHABET



... HELP ...

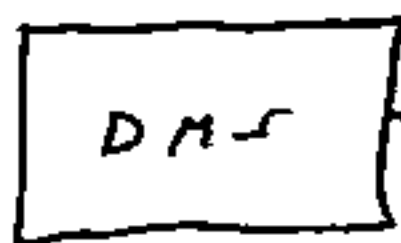
$N = 26$  POSSIBLE  
CHARACTERS OR  
SYMBOLS

$$Q = \{A, B, \dots, Z\} = \{a_1, a_2, \dots, a_N\}$$

$$\text{WITH } P[X_i = a_n] = p_n \quad \text{ALL } i$$

$Q$  IS CALLED THE ALPHABET

EXAMPLE : COIN TOSSING



... 01101 ...

1  $\Rightarrow$  HEADS

0  $\Rightarrow$  TAILS

$$\text{HERE } Q = \{0, 1\} \quad P[X_i = 1] = p = p_1$$

$$P[X_i = 0] = 1 - p = p_2$$

COIN TOSSES ARE INDEPENDENT

ALSO, CALLED A BINARY SOURCE

(JUST A BERNOULLI RANDOM PROCESS!)

MEASURE OF INFORMATION

NEED QUANTITATIVE MEASURE (NOT VALUE OF INFORMATION)

EXAMPLE : EVENT = "IT RAINED YESTERDAY"

⇒ SAME INFORMATION FOR FARMER WHO MUST DECIDE IF CROPS SHOULD BE WATERED OR FOR FACTORY WORKER WHO MUST DECIDE WHETHER <sup>TO</sup> EAT HIS LUNCH OUTSIDE

INTUITIVE PROPERTIES - INFORMATION SHOULD DEPEND ON HOW PROBABLE EVENT IS

EXAMPLE - 1) "CONGRESSMAN INDICTED FOR FRAUD" VS  
2) "CANCER CURE DISCOVERED"

INFO. 2 >> INFO. 1

ONCE REVEALED LESS LIKELY EVENTS CONVEY MORE INFORMATION

ANOTHER PROPERTY IS THAT REVEALING TWO INDEPENDENT EVENT OUTCOMES YIELDS SUM OF INFORMATIONS.

EXAMPLE - 1) "LOC NESS MONSTER FOUND"  
 + "URI GOES TO NCAA FINALS"  
 VS 2) "CURE FOR PROSTRATE CANCER  
 DISCOVERED" + "CURE  
 FOR COLON CANCER DISCOVERED"

INFO<sub>1</sub> = SUM BUT NOT INFO<sub>2</sub>

ALSO, INFORMATION SHOULD BE POSITIVE  
 (CAN'T CANCEL OUT FOR INDEPENDENT  
 EVENTS)

NOW RETURN TO DMS - HOW MUCH  
 INFORMATION IS BEING GENERATED?

$$A = \{a_1, a_2, \dots, a_n\} \quad P\{X_i = a_n\} = p_n$$

A MEASURE OF INFORMATION THAT HAS  
 ALL THE INTUITIVE AND SOME IMPORTANT  
 MATHEMATICAL PROPERTIES IS

$$I(p_j) = \log \frac{1}{p_j}$$

DEPENDS ON PROBS. OF EVENTS ONLY,  
 NOT VALUES, i.e. NOT DEPENDENT ON  $a_j$

FOR  $0 < p_j \leq 1$

- 1)  $I(p_j) \geq 0$
- 2) AS  $p_j$  DECREASES,  $I(p_j)$  INCREASES
- 3) FOR JOINT EVENTS (TWO OUTCOMES)  
THE  $j^{\text{th}}$  OUTCOME IS  $x_j = (x_{j1}, x_{j2})$   
WITH  $P(x_j = a_j) = P(x_{j1} = a_{j1}, x_{j2} = a_{j2})$ .  
IF INDEPENDENT

$$P(x_j = a_j) = P(x_{j1} = a_{j1}) P(x_{j2} = a_{j2})$$

$$p_j = p_{j1} p_{j2}$$

$$\Rightarrow I(p_j) = \log \frac{1}{p_j} = \log \frac{1}{p_{j1} p_{j2}}$$

$$= \log \frac{1}{p_{j1}} + \log \frac{1}{p_{j2}}$$

$$= I(p_{j1}) + I(p_{j2})$$

$\Rightarrow$  INFO. IS ADDITIVE FOR INDEPENDENT EVENTS

$I(p_j)$  CALLED THE SELF-INFORMATION  
USUALLY MEASURED IN BITS PER SOURCE SYMBOL

$$I(p_j) = \log_2 \frac{1}{p_j}$$

OF INFO.  
= NUMBER OF BITS <sub>1</sub> OBTAINED WHEN

WE OBSERVE THAT  $X = a_j$ .

EXAMPLE (\*)  $A = \{A, B, C, D\}$  EQUALLY LIKELY  
 $p_j = 1/4$

$$I(p_j) = \log_2 \frac{1}{p_j} = 2 \text{ BITS/SYMBOL}$$

$j = 1, 2, 3, 4$

EXAMPLE (\*\*)  $A = \{A, B, C, D\}$

$$p_1 = \frac{1}{2}, p_2 = \frac{1}{4}, p_3 = \frac{1}{8}, p_4 = \frac{1}{8}$$

$$I(p_j) = \begin{matrix} 1 & j = 1 \\ 2 & j = 2 \\ 3 & j = 3, 4 \end{matrix}$$

SINCE A DMS WILL EMIT ALL SYMBOLS  
AN AVERAGE MEASURE OF INFO. IS

$$\begin{aligned} E[I(p)] &= \sum_{i=1}^N I(p_i) p_i \\ &= \sum_{i=1}^N p_i \log_2 \frac{1}{p_i} \\ &= - \sum_{i=1}^N p_i \log_2 p_i \text{ BITS/SYMBOL} \\ &= H(X) \text{ SLIGHT ABUSE OF NOTATION} \end{aligned}$$

THIS INFO. CONTENT OF SOURCE IS  
CALLED THE ENTROPY. (UNCERTAINTY OR  
EQUIVOCATION)

EXAMPLE (\*):  $I(p_j) = 2 \quad j = 1, 2, 3, 4$

$\Rightarrow H(x) = E[I(p_j)] = 2$

EXAMPLE (\*\*):  $I(p_j) = \begin{matrix} 1 & j=1 \\ 2 & j=2 \\ 3 & j=3 \\ 3 & j=4 \end{matrix}$

$H(x) = 1 \cdot \frac{1}{2} + 2 \cdot \frac{1}{4} + 3 \cdot \frac{1}{8} + 3 \cdot \frac{1}{8}$   
 $= 1 \frac{3}{4} < 2 \quad \text{WHY?}$

---

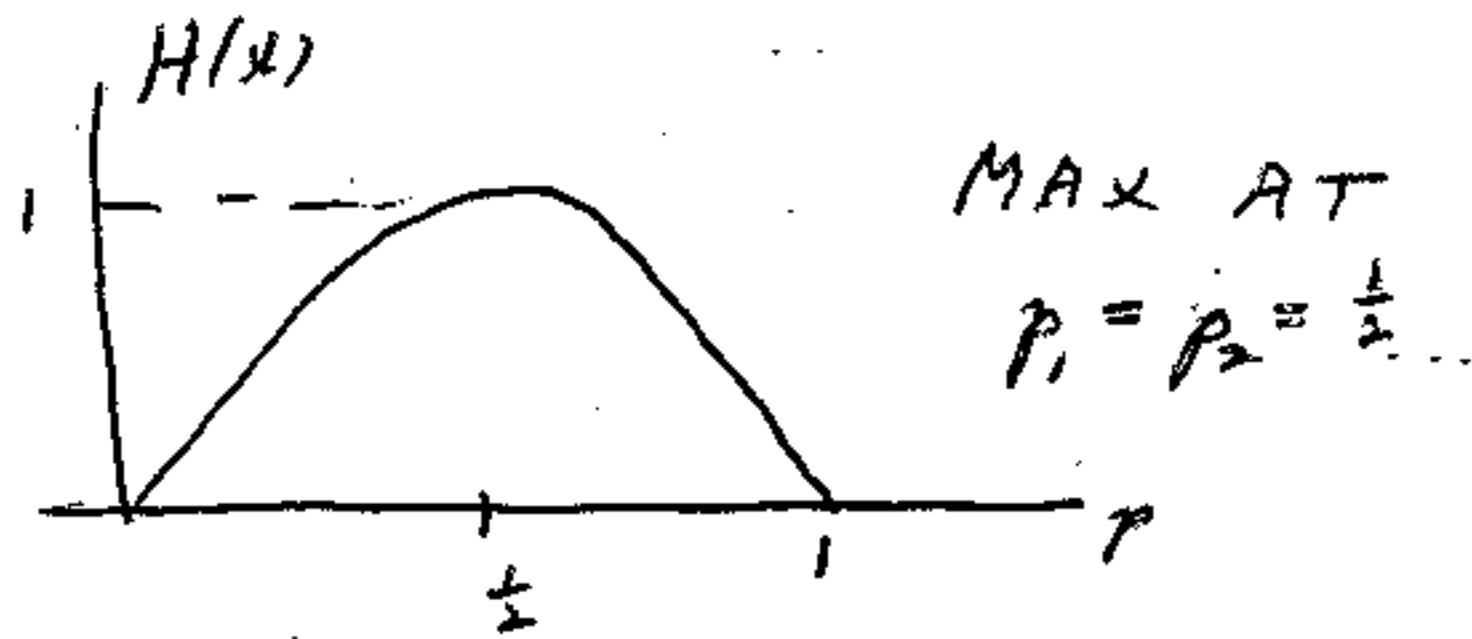
$H(x) = - \sum_{i=1}^N p_i \log_2 p_i \geq 0 \quad \text{WHY?}$

(DEFINE  $0 \log_2 0 = 0$ , EVENT HAS PROB. = 0)

EXAMPLE: BINARY DMS. ( $N=2$ )

$p_1 = p, \quad p_2 = 1-p$

$H(x) = -p \log_2 p - (1-p) \log_2 (1-p)$



NOTE:  $H(x) \Big|_{p=0} = H(x) \Big|_{p=1} = 0 \quad \text{WHY?}$

EXAMPLE: ANALOG  
 A SOURCE HAS  $W = 4000$  Hz  
 SAMPLE AT  $f_s = 2W$   
 QUANTIZE TO  $Q = \{-2, -1, 0, 1, 2\}$   
 WITH  $p_1 = \frac{1}{2}, p_2 = \frac{1}{4}, p_3 = \frac{1}{8}$   
 $p_4 = \frac{1}{16}, p_5 = \frac{1}{16}$

WHAT IS INFORMATION RATE OF SOURCE?

$$H(X) = \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \frac{1}{8} \log_2 8 + \frac{1}{16} \log_2 16 + \frac{1}{16} \log_2 16 = 1.575 \text{ BITS/SAMPLE}$$

$$\text{INFO. RATE} = (\text{BITS/SAMPLE}) \times (\text{SAMPLES/SEC})$$

$$= H(X) f_s = (1.575) (8000)$$

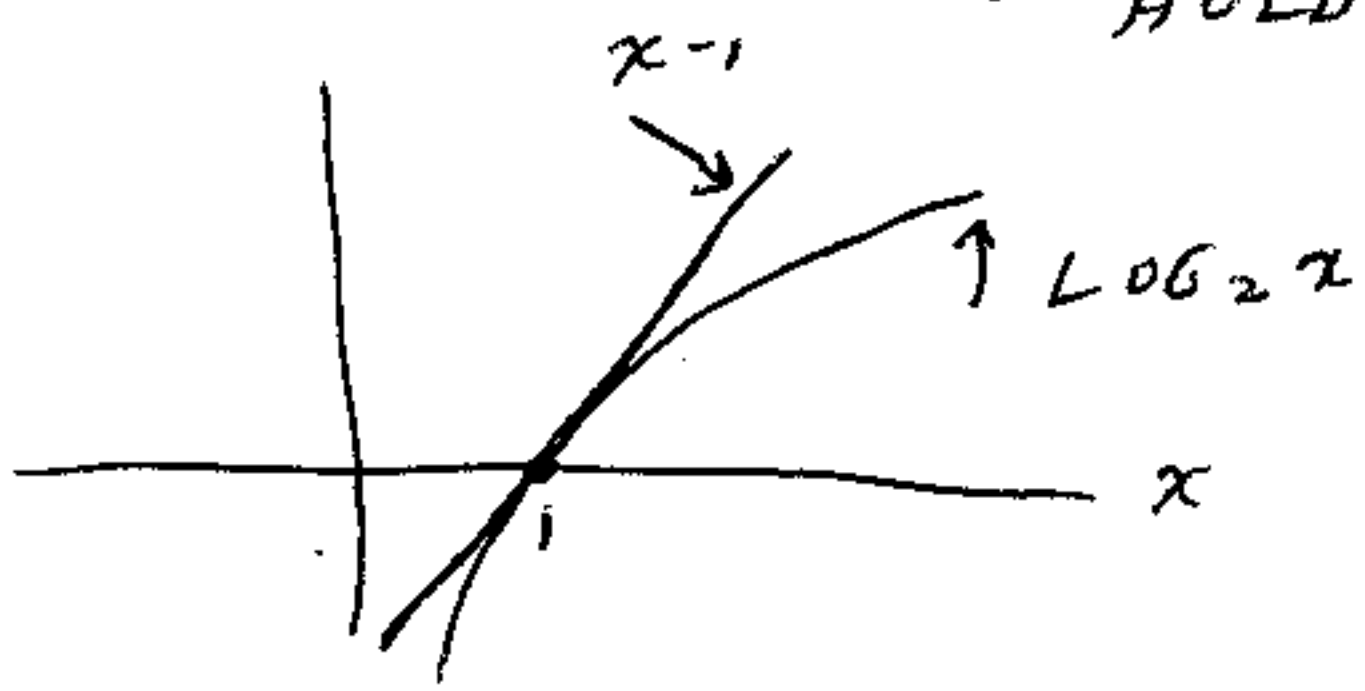
$$= 12,600 \text{ BITS/SEC}$$

ENTROPY IS MAXIMUM FOR  $p_j = 1/N$

$$H(X) = \sum_{j=1}^N p_j \log_2 \frac{1}{p_j} = \log_2 N \sum_{j=1}^N p_j$$

$$= \log_2 N \text{ BITS/SYMBOL}$$

PROOF: NEED  $\log_2 x \leq x-1 \dots x > 0$   
 = HOLDS IFF  $x=1$



$$\begin{aligned}
 H(X) - \log_2 N &= \sum_{j=1}^N p_j \log_2 \frac{1}{p_j} - \log_2 N \\
 &= \sum_{j=1}^N p_j (\log_2 \frac{1}{p_j} - \log_2 N) \\
 &= \sum_{j=1}^N p_j \log_2 \frac{1}{N p_j} \leq \sum_{j=1}^N p_j (\frac{1}{N p_j} - 1) \\
 &= \sum_{j=1}^N (\frac{1}{N} - p_j) = 1 - \sum_{j=1}^N p_j = 0
 \end{aligned}$$

( = for  $\frac{1}{N p_j} = 1$   
 $\Rightarrow p_j = \frac{1}{N}$  )

### SOURCE CODING THEOREM

DUE TO SHANNON

FOR SOURCE WITH ENTROPY  $H(X)$  WE CAN ENCODE SYMBOLS SO THAT ON THE AVERAGE THE CODE WORD LENGTH IS  $H(X)$  BITS/SYMBOL. (BUT HOW?)

THUS, WE CAN COMPRESS SOURCE OUTPUT WITH ALPHABET SIZE  $2^k$  ( $k$  BITS NORMALLY REQUIRED) DOWN TO  $H(X)$ .

OBSVIOUSLY SINCE FOR  $p_j = \frac{1}{2^k}$   $j = 1, 2, \dots, 2^k$  (EQUIPROBABLE OUTPUTS)

$$H(X) = \sum_{j=1}^{2^k} p_j \log_2 \frac{1}{p_j} = \log_2 \frac{1}{\frac{1}{2^k}} = k \text{ BITS / SYMBOL}$$

NO COMPRESSION POSSIBLE.

EXAMPLE :  $N = 5$  SYMBOLS

$\Rightarrow$  NEED  $\log_2 N = \log_2 5$  BITS/SYMBOL  
NEED 3 BITS/SYMBOL

ASSUME  $p_1 = \frac{1}{2}, p_2 = \frac{1}{4}, p_3 = \frac{1}{8}$   
 $p_4 = \frac{1}{16}, p_5 = \frac{1}{16}$

$H(X) = 15/8$  BITS/SYMBOL

HOW DO WE ATTAIN THIS (ON THE AVERAGE)?

		$l$	
$a_1$	-	0	1
$a_2$	-	10	2
$a_3$	-	110	3
$a_4$	-	1110	4
$a_5$	-	1111	4

VARIABLE LENGTH CODE  
 $l$  = CODE WORD LENGTH  
 (RECALL MORSE CODE  
 USE "." FOR "E"  
 "... " FOR "S")

$$E[L] = \sum_{j=1}^5 l_j p_j = 1 \cdot \frac{1}{2} + 2 \cdot \frac{1}{4} + 3 \cdot \frac{1}{8} + 4 \cdot \frac{1}{16} + 4 \cdot \frac{1}{16} = 15/8$$

CHOOSE  $l_j = \log_2 (1/p_j)$  !!

EXPLANATION OF WHY SOURCE CODING THEOREM IS VALID

CONSIDER DMS FOR  $n$  OUTPUTS AND ASK

THE QUESTION: DO WE REALLY NEED TO CONSIDER TRANSMITTING ONE OF  $N^n$  POSSIBLE SEQUENCES ( $N =$  ALPHABET SIZE)?

ANSWER: NO, IF SOME ALPHABET OUTPUTS ARE MORE PROBABLE THAN OTHERS.

ASSUME  $n$  IS LARGE  $\Rightarrow$  OUTPUT SEQUENCE HAS  $(np_1)$   $a_1$ 'S,  $(np_2)$   $a_2$ 'S, ...,  $(np_N)$   $a_N$ 'S  
(LAW OF LARGE NUMBERS - RECALL COIN TOSS WITH  $p = P(\text{HEADS}) = 3/4 \Rightarrow$   
... HTHHTHHHTHHHT...  $\approx 3/4 n$  HEADS AND  $1/4 n$  TAILS)

A TYPICAL SEQUENCE  $x$  ( $3/4 n$  HEADS,  $1/4 n$  TAILS FOR EXAMPLE)  
HAS PROB. APPROXIMATELY

$$P(x = \underline{x}) \approx \prod_{i=1}^N p_i^{np_i}$$

$$\left( \begin{array}{l} \text{HHHTHHHT} \Rightarrow P = p^6 (1-p)^2 \\ N=2 \\ n=8 \end{array} \right) = p^{8(3/4)} (1-p)^{8(1/4)}$$

$$= \prod_{i=1}^N 2^{-\log_2 p_i^{np_i}} = \prod_{i=1}^N 2^{-np_i \log_2 p_i}$$

$$= 2^{\sum_{i=1}^N n p_i \log_2 p_i} = 2^{-n(-\sum p_i \log_2 p_i)}$$

$$= 2^{-nH(X)}$$

TYPICAL

SINCE ALL SEQUENCES ARE EQUALLY LIKELY AND ATYPICAL SEQUENCES HAVE PROB.  $\rightarrow 0$  AS  $n \rightarrow \infty$ , THERE MUST BE

$$2^{nH(X)} \text{ TYPICAL SEQUENCES}$$

$\Rightarrow$  NEED  $nH(X)$  BITS TO ENCODE THE TYPICAL  $n$ -LENGTH SEQUENCES  $\Rightarrow$  NEED

$H(X)$  BITS/SYMBOL !

THIS IS AN EXISTENCE PROOF - DOESN'T SAY HOW TO DO IT. PROVIDES THE HINT THAT WE MUST CODE NOT INDIVIDUAL SYMBOLS BUT INSTEAD LONG BLOCKS OF SYMBOLS.

HUFFMAN CODING

ATTAINS  $H(X)$  BITS/SYMBOL FOR  $n$  LENGTH SEQUENCE AS  $n \rightarrow \infty$ , USES VARIABLE LENGTH CODE WORDS.

## TYPES OF VARIABLE LENGTH CODES

Letter	Probability	Codewords			
		Code 1	Code 2	Code 3	Code 4
$a_1$	$p_1 = \frac{1}{2}$	1	1	0	00
$a_2$	$p_2 = \frac{1}{4}$	01	10	10	01
$a_3$	$p_3 = \frac{1}{8}$	001	100	110	10
$a_4$	$p_4 = \frac{1}{16}$	0001	1000	1110	11
$a_5$	$p_5 = \frac{1}{16}$	00001	10000	1111	110

CODE 4 IS NO GOOD SINCE

$110110 \rightarrow 110, 110 \Rightarrow a_5, a_5$

OR  $11, 01, 10 \Rightarrow a_4, a_2, a_3$

NOT UNIQUELY DECRYPTABLE (DON'T WANT TO INSERT BLANKS, SLOWS TRANSMISSION RATE)

CODE 1, 2 ARE O.K. BUT

CODE 1 ... 001 001 01 ...  
 $a_3 \quad \uparrow \quad a_3 \quad a_2$

MAKE DECISION

CODE 2 ... 100 1000 100 ...

$a_3 \quad \uparrow$  MAKE DECISION - MUST WAIT ONE BIT TO SEE IF IT

IS A 1

CODE 1 IS INSTANTANEOUS BUT NOT CODE 2

CODES 1, 3 ARE BOTH UNIQUELY DECODABLE AND INSTANTANEOUS BUT

$$E(L) = \frac{31}{16}$$

$$\frac{30}{16}$$

CODE 1  
CODE 3

CODE 3 IS HUFFMAN CODE - LEAST AVERAGE  
AVERAGE  
CODEWORD LENGTH (CAN BE NONINTEGER  
NUMBER OF BITS) - FOR LONG STRINGS  
OF SYMBOLS IT REQUIRES LEAST NUMBER  
OF BITS

EXAMPLE

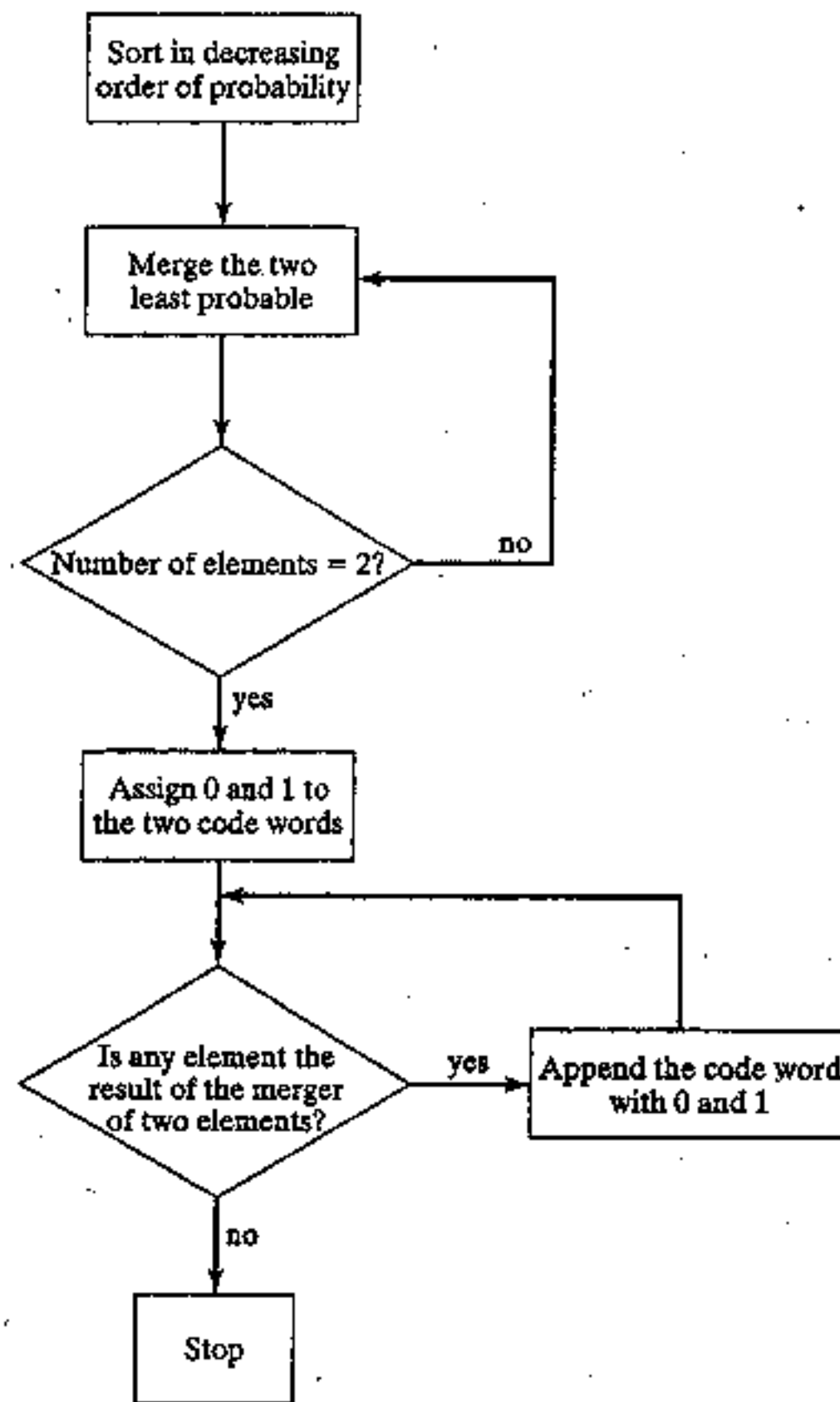
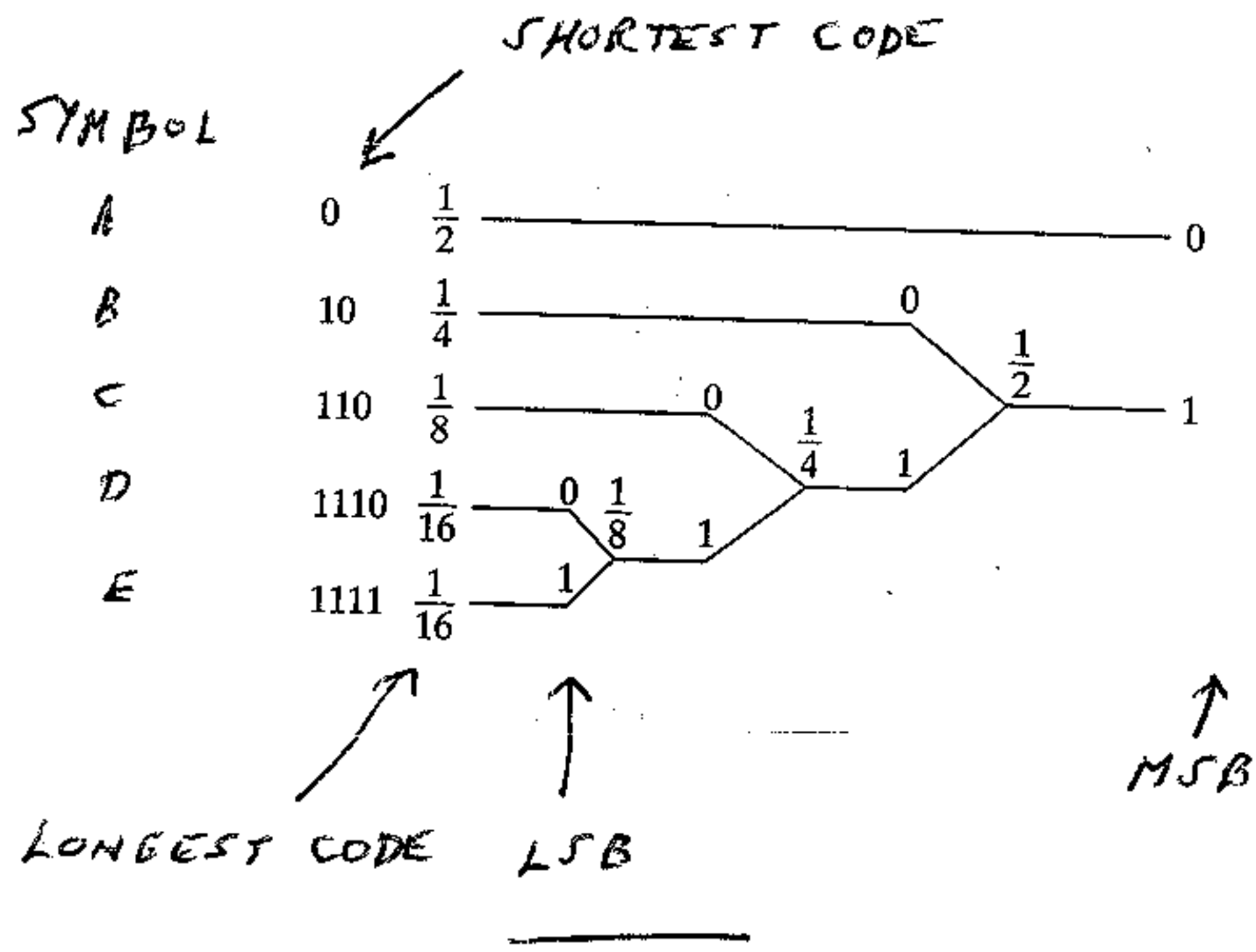


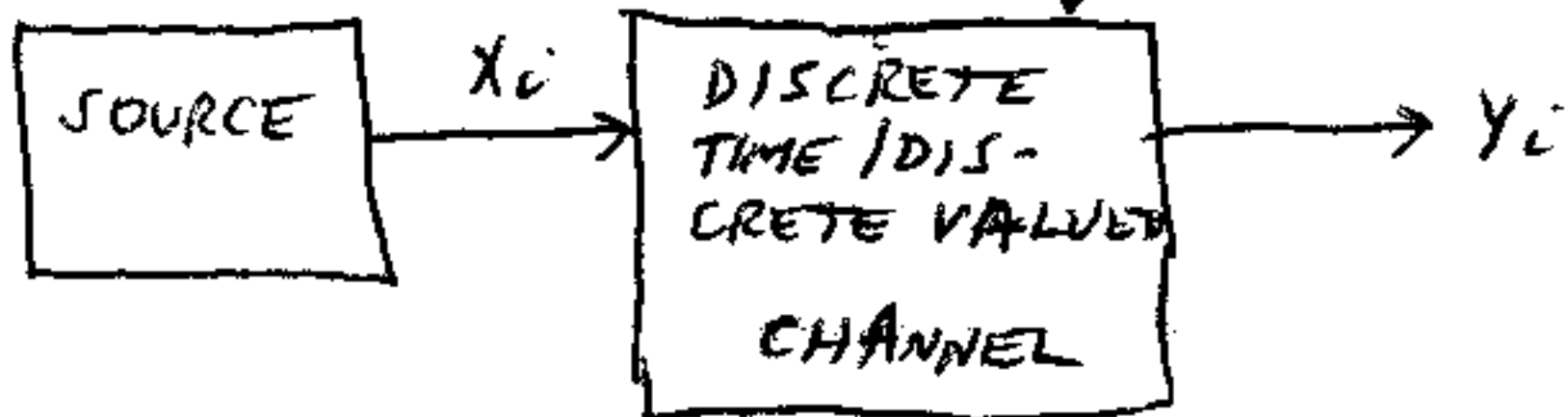
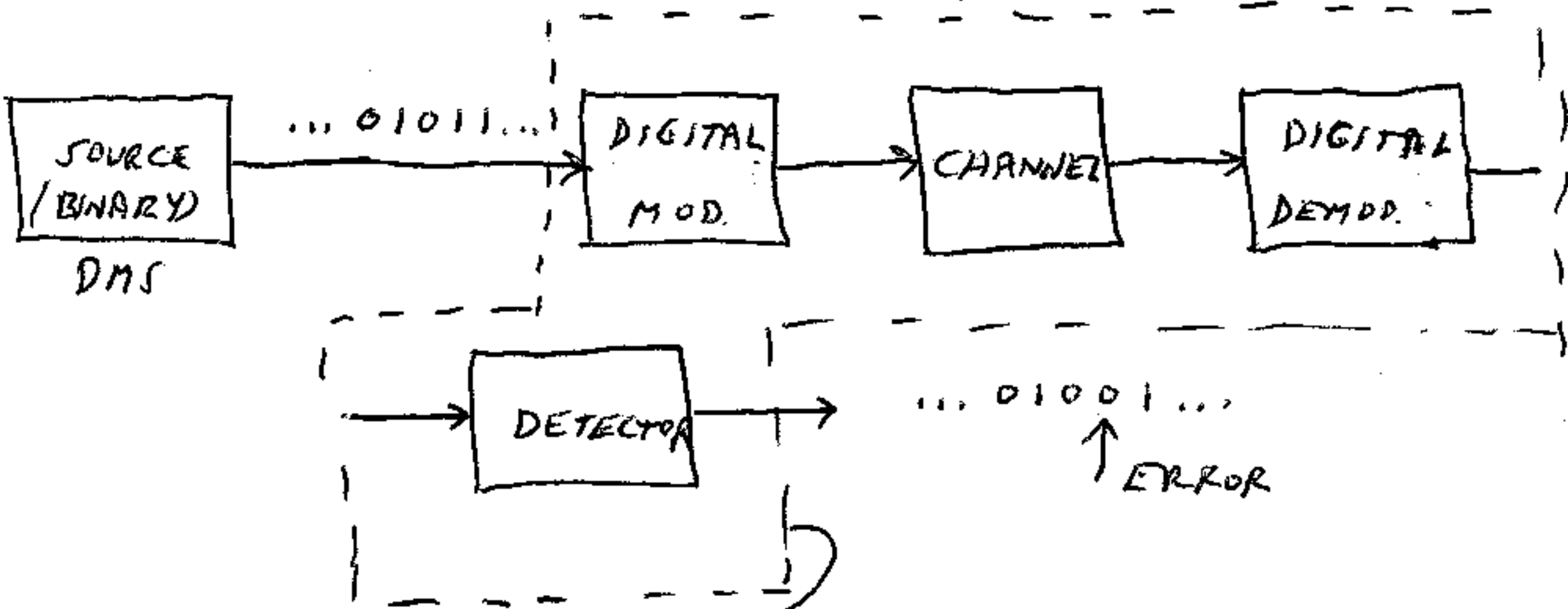
Figure 12.6 Huffman coding algorithm.

$P_A = \frac{1}{2}, P_B = \frac{1}{4}, P_C = \frac{1}{8}, P_D = \frac{1}{16}, P_E = \frac{1}{16}$



THIS WORKS PERFECTLY, (i.e.  $E(L) = H(X)$ )  
 SINCE  $\log_2 \frac{1}{p} = \text{INTEGER}$ . IF THIS IS  
 NOT THE CASE WE MUST ENCODE BLOCKS  
 OF SYMBOLS TO HAVE  $E(L)$  NEAR TO  
 $H(X)$ . SEE EXAMPLE 12.3.3.

MODELING OF COMM. CHANNELS



$x_i = i^{th}$  INPUT BIT  
 $y_i = i^{th}$  OUTPUT BIT

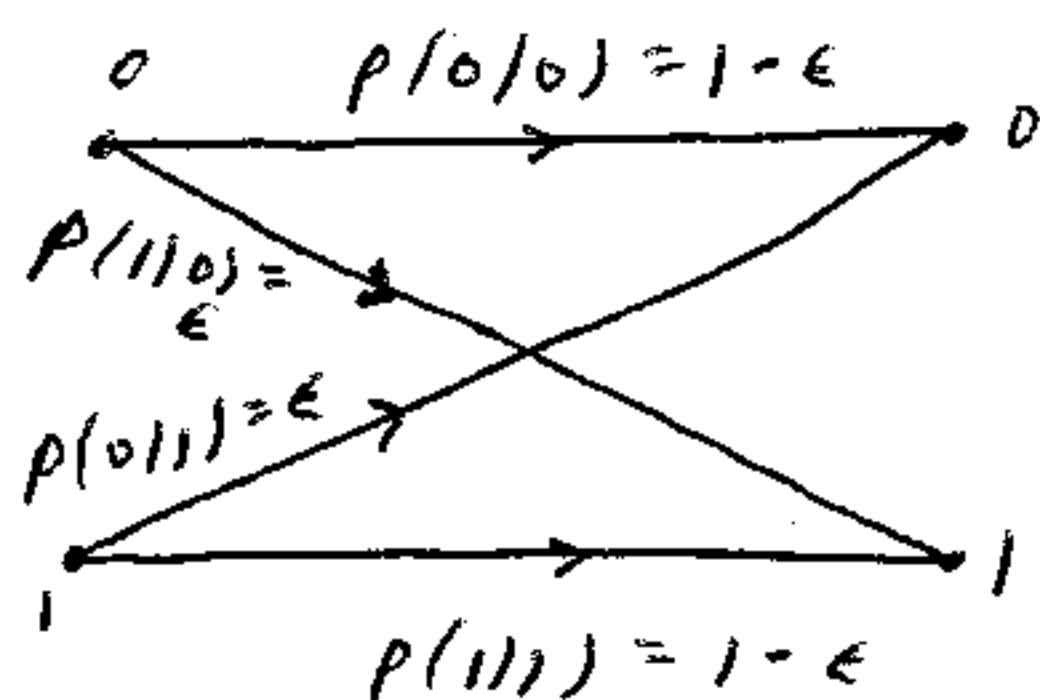
TO CHARACTERIZE DTDV CHANNEL JUST NEED  $P(y_i | x_i)$  FOR ALL VALUES OF  $(x_i, y_i)$ . HERE WE HAVE

$$x_i = 0, 1$$

$$y_i = 0, 1$$

$$\Rightarrow \underbrace{P(010), P(110)}_{\text{SUM TO 1}}, \underbrace{P(011), P(111)}_{\text{SUM TO 1}}$$

EXAMPLE : BINARY SYMMETRIC CHANNEL



SYMMETRIC SINCE  
 $P(1/0) = P(0/1)$   
 $= \epsilon$

THIS MODELS A BPSK SYSTEM (ANTIPODAL SIGNALS) WITH

$$\epsilon = P(1/0) = P(0/1) = Q\left(\sqrt{\frac{2E_b}{N_0}}\right)$$

NOTE: WE HAVE IMPLICITLY ASSUMED THAT EFFECT OF CHANNEL ON BIT  $i$  DOES NOT DEPEND ON BIT  $j$ . (NO INTERSYMBOL INTERFERENCE FOR EXAMPLE.  $\Rightarrow$ )

DISCRETE MEMORYLESS CHANNEL (DMC)

CHANNEL CAPACITY

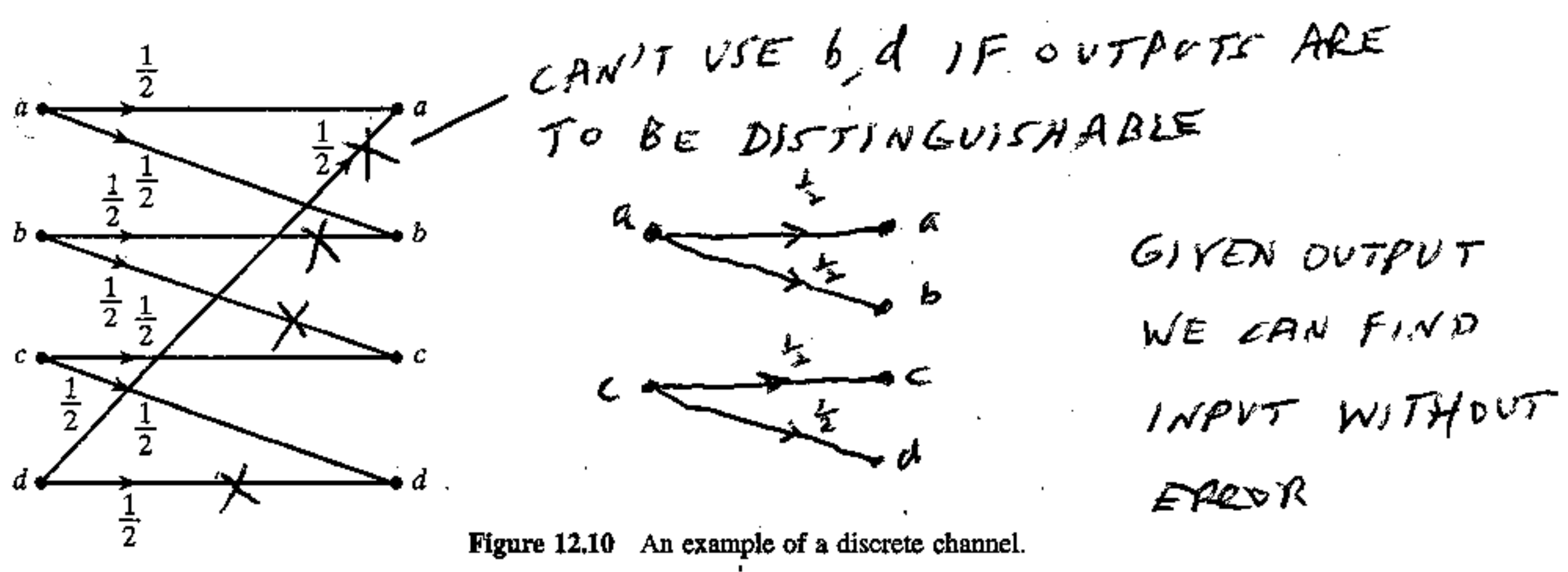
IN 1948, CLAUDE SHANNON PROVED THAT ERRORLESS TRANSMISSION IS POSSIBLE AS LONG AS TRANSMISSION RATE  $R$  BPS IS LESS THAN THE CHANNEL CAPACITY  $C$ .

THIS IS NOISY CHANNEL CODING THEOREM

PREVIOUSLY THIS WAS THOUGHT TO BE IMPOSSIBLE!

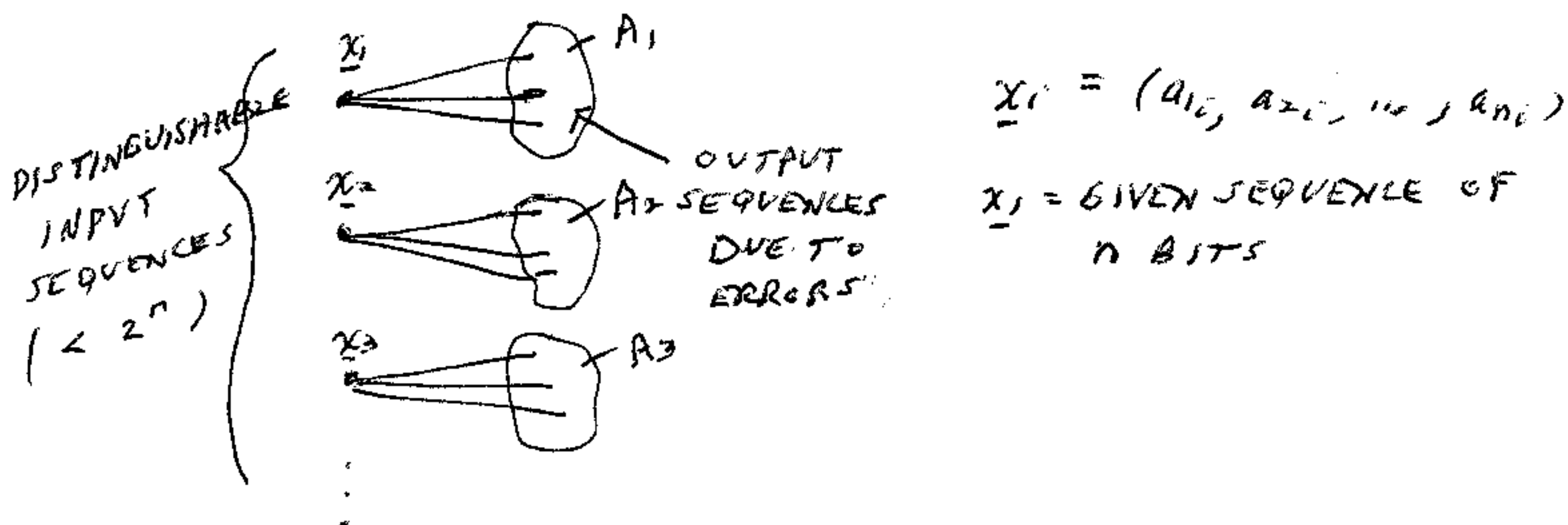
C IS A FUNDAMENTAL LIMIT ON TRANSMISSION SPEED FOR  $P_e \rightarrow 0$ .

KEY IDEA: FOR ERRORLESS TRANSMISSION POSSIBLE OUTPUTS OF CHANNEL MUST BE DISTINGUISHABLE.



USE ONLY INPUTS THAT RESULT IN <sup>BINARY</sup> DISJOINT OUTPUTS. FOR A <sup>n</sup> DAC WE CAN ONLY DISTINGUISH A SUBSET OF THE  $2^n$  <sup>OUTPUT</sup> SEQUENCES THAT ARE POSSIBLE FOR AN INPUT SEQUENCE OF n BITS.

HOW MANY SEQUENCES ARE IN EACH DISJOINT SUBSET  $A_i$  AS SHOWN?



IF THIS PICTURE HOLDS, WE CAN DECODE  
WITH NO ERROR. (IF  $\underline{x} \in A_i \Rightarrow \underline{x}_i$  TRANSMITTED)

FOR EACH  $\underline{x}_i$  TRANSMITTED WE WILL  
HAVE  $n\epsilon$  ERRORS (WITH PROB. 1 AS  $n \rightarrow \infty$ )

$\Rightarrow n\epsilon$  BITS WILL BE DIFFERENT THAN  $\underline{x}_i$

$\binom{n}{n\epsilon}$  POSSIBLE OUTPUTS FOR EACH  $\underline{x}_i$

EXAMPLE:  $n=4, \epsilon = 1/4 \Rightarrow n\epsilon = 1$  ERROR

TRANSMIT 0110 BUT RECEIVE

1110, 0010, 0100, 0111 =  $\binom{4}{1}$   
SEQUENCES

$\Rightarrow$  SIZE OF  $A_i = \binom{n}{n\epsilon}$

SIZE OF  $A_1 \cup A_2 \cup A_3 \cup \dots =$  TOTAL  
POSSIBLE OUTPUT SEQUENCES

$$= 2^{nH(Y)} \quad (\text{TYPICAL ONES})$$

FINALLY NUMBER OF DISTINGUISHABLE  
INPUT SEQUENCES IS

$$M = \frac{2^{nH(Y)}}{\binom{n}{n\epsilon}} = \frac{\# \text{ TYPICAL OUTPUT SEQUENCES}}{\# \text{ SEQUENCES IN EACH } A_\epsilon \text{ SET}}$$

$$\text{NOW } \binom{n}{n\epsilon} = \frac{n!}{(n-n\epsilon)!(n\epsilon)!}$$

$$n! \approx \sqrt{2\pi n} n^n e^{-n} \quad (\text{STIRLING APPROX.})$$

$$\begin{aligned} \binom{n}{n\epsilon} &\approx \frac{\sqrt{2\pi n} n^n e^{-n}}{\sqrt{2\pi n(1-\epsilon)} [n(1-\epsilon)]^{n(1-\epsilon)} e^{-n(1-\epsilon)}} \\ &\quad \cdot \frac{1}{\sqrt{2\pi n\epsilon} [n\epsilon]^{n\epsilon} e^{-n\epsilon}} \\ &= \frac{n^n}{[n(1-\epsilon)]^{n(1-\epsilon)} (n\epsilon)^{n\epsilon}} \cdot \frac{1}{\sqrt{1-\epsilon} \sqrt{2\pi n\epsilon}} \end{aligned}$$

$$\begin{aligned} \log_2 \binom{n}{n\epsilon} &\approx n \log_2 n - n(1-\epsilon) \log_2 n - n(1-\epsilon) \log_2 (1-\epsilon) \\ &\quad - n\epsilon \log_2 n\epsilon - \frac{1}{2} \log_2 (1-\epsilon) - \frac{1}{2} \log_2 2\pi \\ &\quad - \frac{1}{2} \log_2 n\epsilon \quad \leftarrow \begin{array}{l} \nwarrow \\ \nearrow \end{array} \begin{array}{l} \text{TERMS ARE SMALL} \\ \text{OTHERS ARE } O(n) \end{array} \\ &= +n\epsilon \log_2 n - n(1-\epsilon) \log_2 (1-\epsilon) - n\epsilon \log_2 n\epsilon \\ &= -n(1-\epsilon) \log_2 (1-\epsilon) - n\epsilon \log_2 \epsilon = nH_b(\epsilon) \end{aligned}$$

WHERE  $H_b(\epsilon) = -\epsilon \log_2 \epsilon - (1-\epsilon) \log_2 (1-\epsilon)$

$$\Rightarrow M = \frac{2^{nH(Y)}}{2^{nH_b(\epsilon)}} = 2^{n(H(Y) - H_b(\epsilon))}$$

THUS, WE CAN ONLY TRANSMIT AT MOST  $M$  MESSAGES,  $\{x_1, x_2, \dots, x_M\}$  USING  $n$  BITS.  
OR

$\log_2 M$  INFORMATION BITS PER  $n$  CHANNEL USES OR

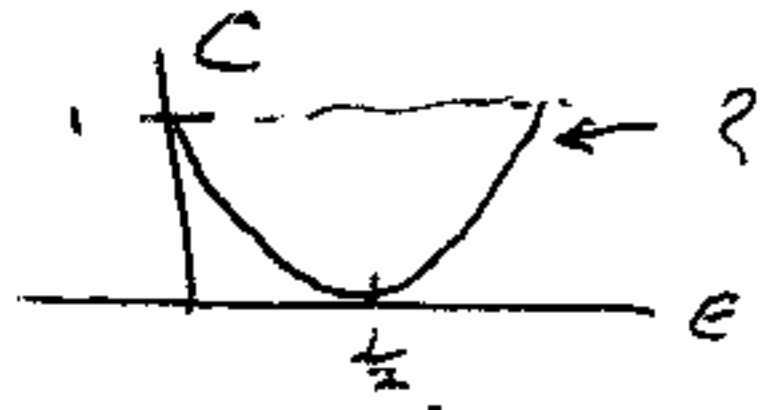
$$\frac{\log_2 M}{n} = R = H(Y) - H_b(\epsilon)$$

↑  
TRANSMISSION RATE

FINALLY, SINCE WE MAY BE ABLE TO INCREASE  $R$  BY ADJUSTING  $H(Y)$  (HAVE TO MODIFY  $P(0), P(1)$  AT INPUT, AND  $H(Y) \leq 1$  (BINARY CASE)

$$R_{max} = C = 1 - H_b(\epsilon)$$
$$= \max_{P(0), P(1)} H(Y) - H_b(\epsilon)$$

$C =$  CHANNEL CAPACITY



MEASURED IN INFORMATION BITS PER CHANNEL USE. CHANNEL IS USELESS