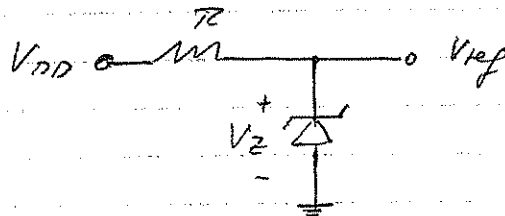


VI Voltage References

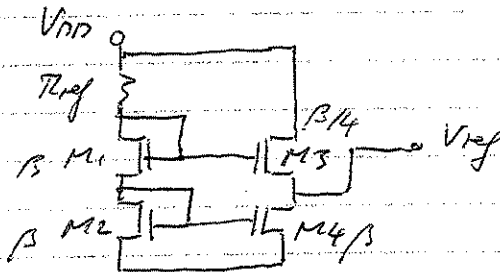
Solution 1 Making use of a Zener diode that breaks down at a specific (reverse) junction voltage

Basic Configuration

$$V_{ref} < V_{DD}$$

$$\| V_{ref} = V_z \|$$

Solution 2 Making use of the Threshold voltage of an MOS device

Basic Configuration

$$V_{ref} = V_{t1} + V_{eff1} + V_{t2} + V_{ref2} - V_{t3} - V_{eff3}$$

if the width of M_3 is about 4 times smaller than the width of the other 3 devices

$$\| V_{ref} \approx V_{t1} + V_{t2} - V_{t3} \approx V_{t2} \|$$

Note: The above relationship holds independent of the supply voltage V_{DD} .

Problem: V_t does vary with temperature.

Temperature Dependence of n_i and Φ_F

$$| n_i = 2 \left(\frac{2\pi m_n^* kT}{h^2} \right)^{3/2} e^{-\frac{E_C - E_F}{kT}} | \quad m_n^* \text{ Effective mass of e.l.}$$

$$| p_i = 2 \left(\frac{2\pi m_p^* kT}{h^2} \right)^{3/2} e^{-\frac{E_F - E_V}{kT}} | \quad m_p^* \text{ Effective mass of hole}$$

Since $n_i = p_i$ (intrinsic semiconductor)

$$| n_i = 2 \left(\frac{2\pi kT}{h^2} \right)^{3/2} (m_n^* m_p^*)^{3/4} e^{-\frac{E_C - E_V}{2kT}} |$$

$$| n_i(T) = n_0 \left(\frac{T}{T_0} \right)^{3/2} e^{-\frac{E_C - E_V}{2kT}} |$$

$$\frac{\partial n_i}{\partial T} = n_0 \left(\frac{T}{T_0} \right)^{3/2} e^{-\frac{E_C - E_V}{2kT}} \left[\frac{3}{2} \frac{1}{T} + \frac{E_C - E_V}{2kT^2} \right]$$

$$\therefore \left| \frac{\partial n_i}{\partial T} = n_i \left[\frac{3}{2} \frac{1}{T} + \frac{E_G}{2kT^2} \right] \right| \quad E_G = E_C - E_V$$

$$| \Phi_F = \frac{kT}{q} \ln \left(\frac{N_{sub}}{n_i} \right) |$$

$$\frac{\partial \Phi_F}{\partial T} = \frac{k}{q} \ln \left(\frac{N_{sub}}{n_i} \right) - \frac{kT}{q} \frac{n_i}{N_{sub}} \frac{N_{sub}}{n_i} \frac{\partial n_i}{\partial T}$$

$$\therefore \left| \frac{\partial \Phi_F}{\partial T} = \frac{V_T}{T} \left[\ln \left(\frac{N_{sub}}{n_i} \right) - \frac{3}{2} - \frac{E_G}{2kT} \right] \right| \quad V_T = \frac{kT}{q}$$

$$@ T = 300^\circ K \text{ and } N_{sub} \approx 10^7 n_i \quad \left| \frac{\partial \Phi_F}{\partial T} \approx -0.5 \text{ mV}/^\circ K \right|$$

Temperature Dependence of V_d and V_t

$$V_d \approx V_T \ln\left(\frac{I_d}{I_s}\right)$$

$$I_s \approx I_{s0} \left(\frac{T}{T_0}\right)^{3/2} e^{-\frac{E_g}{kT}}$$

$$\therefore \left\| \frac{\partial I_s}{\partial T} = I_{s0} \left(\frac{T}{T_0}\right)^{3/2} e^{-\frac{E_g}{kT}} \left[\frac{3}{2} \frac{1}{T} + \frac{E_g}{kT^2} \right] = I_s \left[\frac{3}{2} \frac{1}{T} + \frac{E_g}{kT^2} \right] \right\|$$

$$\frac{\partial V_d}{\partial T} = \frac{V_T}{T} \ln\left(\frac{I_d}{I_s}\right) + V_T \frac{I_s}{I_d} \left[\frac{\partial I_d}{I_s \partial T} - \frac{I_d}{I_s^2} \frac{\partial I_s}{\partial T} \right]$$

$$\therefore \frac{\partial V_d}{\partial T} = \frac{V_d}{T} + V_T \left[\frac{\partial I_d}{I_d \partial T} - \frac{\partial I_s}{I_s \partial T} \right]$$

$$\left\| \frac{\partial V_d}{\partial T} = \frac{V_d}{T} + \frac{V_T}{T} \left[\frac{1}{I_d} \frac{\partial I_d}{\partial T} - \frac{3}{2} - \frac{E_g}{kT} \right] \right\|$$

If $I_d \approx I_{d0} \cdot \frac{T}{T_0}$ and $T = 300^\circ K$

$$\left\| \frac{\partial V_d}{\partial T} = \frac{V_d}{T} - \frac{V_T}{T} \left[\frac{5}{2} + \frac{E_g}{kT} \right] \approx -1.8 \text{ mV}/^\circ K \right\|$$

$$V_{tn} = V_{FB} + 2\phi_n + \frac{1}{C_{ox}} \sqrt{2\phi_n 2\epsilon_s q N_A}$$

$$\therefore \left\| \frac{\partial V_{tn}}{\partial T} \approx \frac{\partial \phi_n}{\partial T} \left[2 + \frac{\sqrt{2\epsilon_s q N_A}}{C_{ox} \sqrt{2\phi_n}} \right] \right\|$$

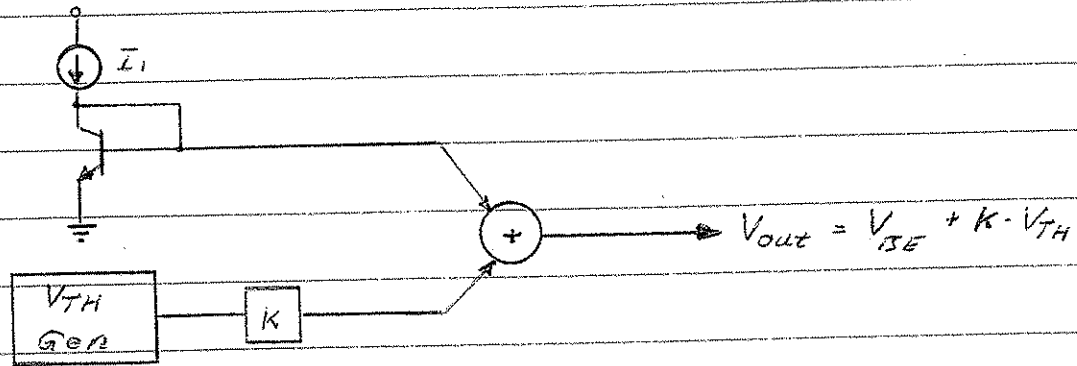
If $C_{ox} = 2 \times 10^{-3} \frac{F}{m^2}$ and $N_A = 10^{25} m^{-3}$

$$\left\| \frac{\partial V_{tn}}{\partial T} \approx -1.5 \text{ mV}/^\circ K \right\| \quad @ T = 300^\circ K$$

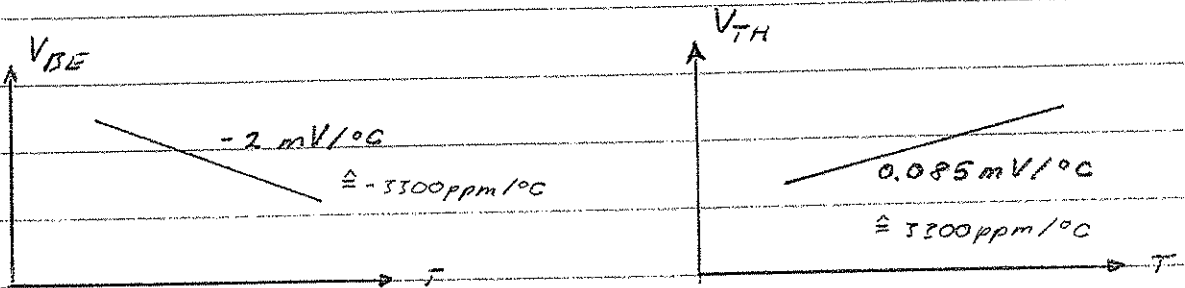
Note: $\frac{\partial V_{tp}}{\partial T} \approx -\frac{\partial V_{tn}}{\partial T}$

3. Temperature independent Biasing (Bandgap Reference)

Principle of Operation



Temp. Coefficients



T_C of $V_{BE}(\text{on})$

$$I_1 \approx I_S e^{V_{BE}/V_{TH}} \Rightarrow V_{BE} \approx V_{TH} \ln\left(\frac{I_1}{I_S}\right)$$

$$\begin{aligned} I_1 &= \text{const} \\ I_1 &= A \cdot T^\alpha \\ I_S &= \text{const} \\ I_S &= B \cdot T^{4-n} e^{-\frac{V_{GO}}{V_{TH}}} \end{aligned}$$

typically $0 < \alpha < 1$

V_{GO} : Bandgap voltage extrapolated to 0°K

n depends on doping level

in Base e.g. $n = 3/2$

VI - 5

$$V_{BE} \approx V_{TH} \ln \left(G \cdot T^\alpha \cdot \beta^{-1} \cdot T^{n-4} e^{\frac{V_{GO}}{V_{TH}}} \right)$$

$$= V_{GO} + V_{TH} \ln \left(\frac{G}{\beta} T^{n-4+\alpha} \right)$$

$$\underline{V_{BE} = V_{GO} - V_{TH} \left([4-n-\alpha] \ln T - \ln \frac{G}{\beta} \right)}$$

output voltage after summer:

$$\| V_{out} = V_{GO} - V_{TH} [4-n-\alpha] \ln T + V_{TH} \left(K + \ln \frac{G}{\beta} \right) \|$$

$$\left. \frac{dV_{out}}{dT} \right|_{T=T_0} = -[4-n-\alpha] [1 + \ln T_0] \frac{V_{TH0}}{T_0} + \frac{V_{TH0}}{T_0} \left(K + \ln \frac{G}{\beta} \right)$$

$$= \frac{V_{TH0}}{T_0} \left(K + \ln \frac{G}{\beta} - [4-n-\alpha] [1 + \ln T_0] \right)$$

$$\underline{\left. \frac{dV_{out}}{dT} \right|_{T=T_0} = 0}$$

$$\Rightarrow \| K + \ln \frac{G}{\beta} = [4-n-\alpha] [1 + \ln T_0] \|$$

↓

$$\| \begin{aligned} V_{out} &= V_{GO} - V_{TH} [4-n-\alpha] \ln T + V_{TH} [4-n-\alpha] [1 + \ln T_0] \\ &= V_{GO} + V_{TH} [4-n-\alpha] \left[1 + \ln \left\{ \frac{T_0}{T} \right\} \right] \end{aligned} \|$$

Note: - Temp dependence of V_{out} is described by T_0
 - T_0 is determined by k, B, G

Example: given: $n = 3/2$; $\alpha = 1$; $V_{A0}(Si) = 1.2V$
 find: $V_{out}|_{T_0=25^\circ C}$ such that $T_c|_{T_0=25^\circ C} = 0$

solution: $T_c = 0$ if

$$V_{out} = V_{A0} + V_{TH} [4 - n - \alpha] [1 + \ln\{\frac{T_0}{T}\}]$$

check: $\frac{dV_{out}}{dT} = [4 - n - \alpha] \left[\underbrace{\frac{V_{TH}}{T} (1 + \ln\{\frac{T_0}{T}\}) - \frac{V_{TH}}{T} \frac{T_0}{T}}_{= 0 \text{ if } T = T_0} \right]$

$$\underline{V_{out} = 1.2V + 0.026V \cdot 1.5 = 1.239V}$$

assume that due to component variations $V_{out} = 1.250V$

find Temp at which $T_c = 0$

and $T_c|_{T=25^\circ C}$

solution: $V_{out} = V_{A0} + V_{TH} [4 - n - \alpha] [1 + \ln\{\frac{T_0}{T}\}] + V_{TH} \Delta$

$$\Delta = \frac{V_{out0} - V_{out}}{V_{TH}} \quad \frac{dV_{out}}{dT} = [4 - n - \alpha] \frac{V_{TH}}{T} \ln\{\frac{T_0}{T}\} + \frac{V_{TH}}{T} \Delta$$

$$\frac{dV_{out}}{dT} = 0 \Rightarrow [4 - n - \alpha] \ln\{\frac{T_0}{T}\} + \Delta = 0$$

$$\underline{T = T_0 \cdot e^{\frac{\Delta}{[4 - n - \alpha]}} = 300^\circ K \cdot e^{\frac{0.42}{1.5}} = 398^\circ K}$$

$$\underline{T_c|_{T=25^\circ C} = \frac{1}{V_{out}} \frac{dV_{out}}{dT} \bigg|_{T=25^\circ C} = \frac{1}{V_{out}} \frac{V_{TH}}{T_0} \Delta = 29 \text{ ppm}/^\circ C$$

B. MOS

1. Circuit

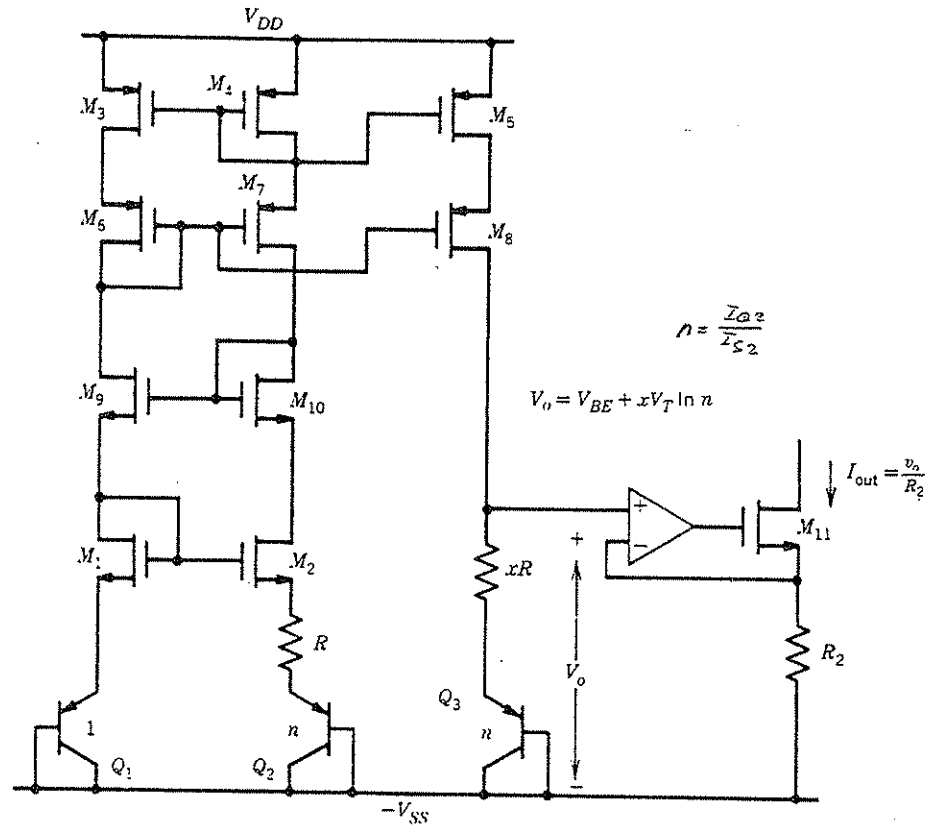


Figure 12.29 Example of a V_{BG} -referenced self-biased reference circuit.

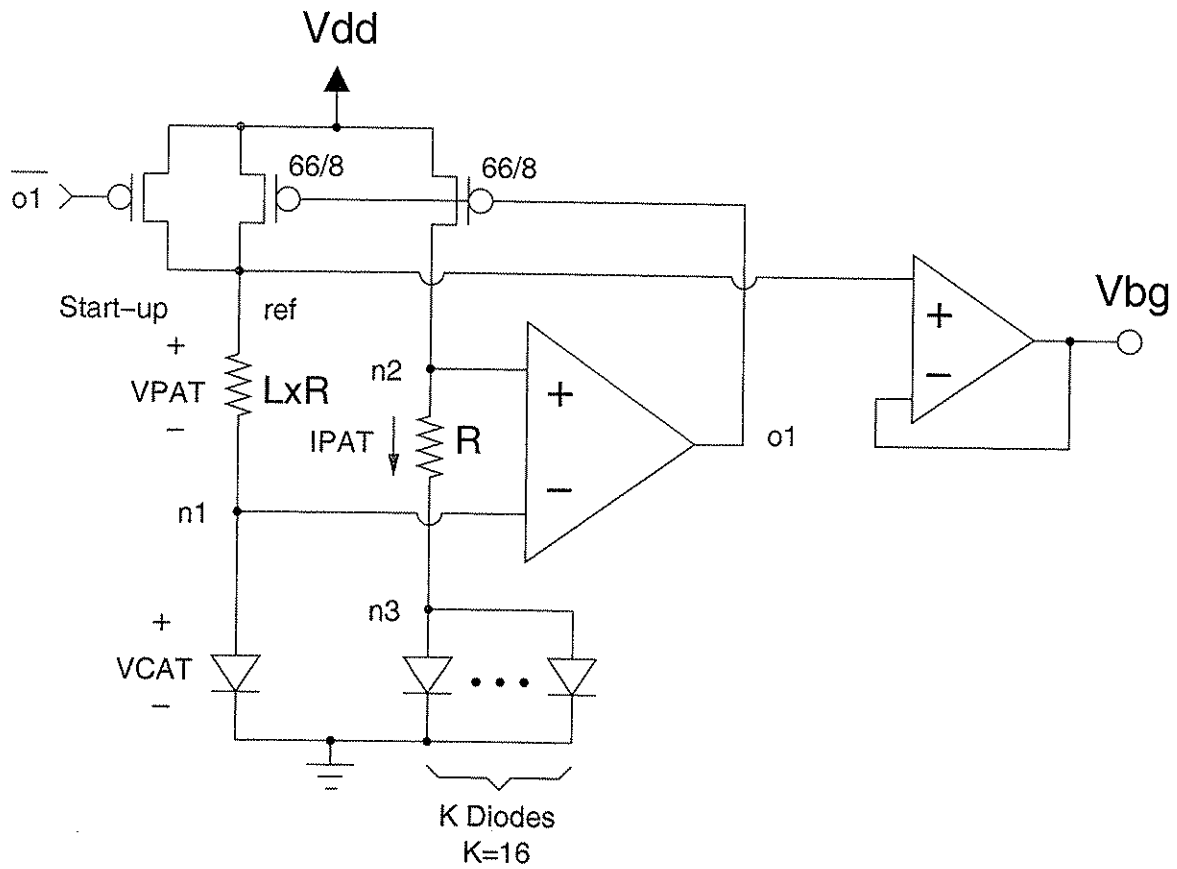
Notes: $Q_1 - Q_3$ can only be implemented with an n-well process since their collectors must be connected to $-V_{SS}$

$$I_{OUT} = V_o/R_2$$

$$V_{OUT} = I_{Q3}xR + V_{BE3}$$

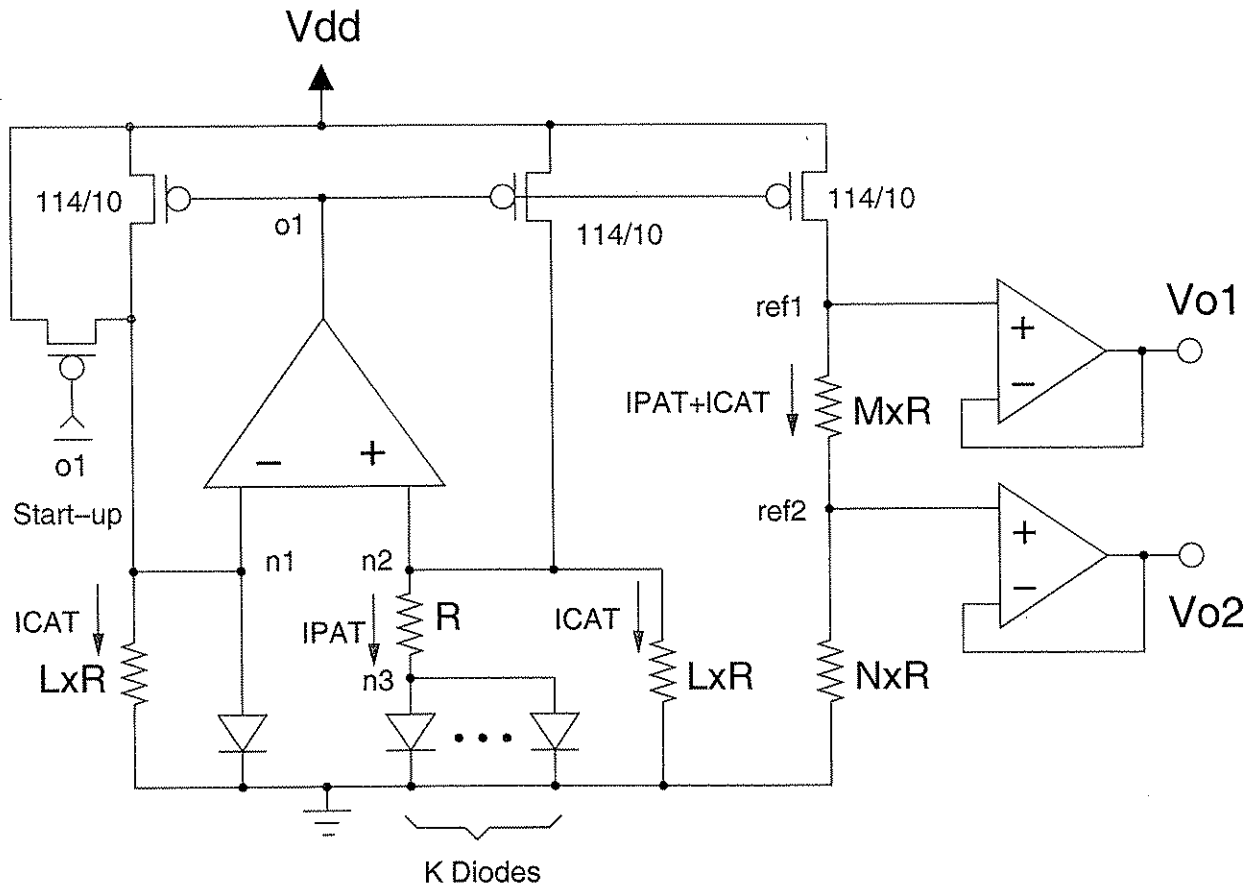
CMOS Bandgap Reference Circuit

Version1: Generating temperature compensated Voltage

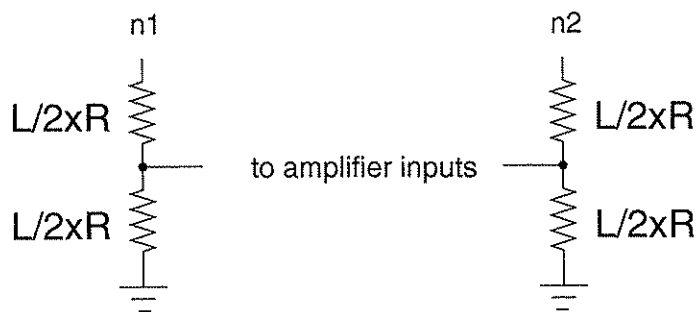


CMOS Bandgap Reference Circuit

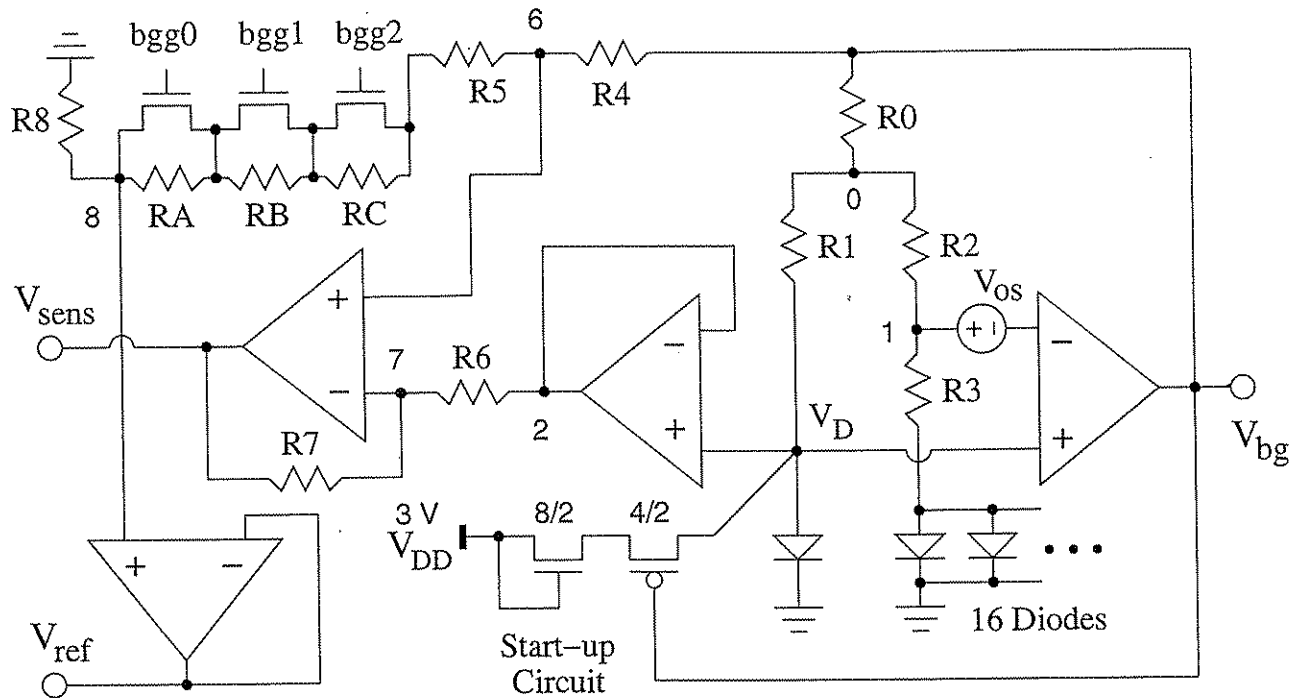
Version2: generating temperature compensated Current



To lower amplifier common mode input replace LxR branches by



Thermal Sensor & Bandgap Reference Circuit



Thermal Stability Condition

$$V_T \frac{R_2}{R_3} \left(1 + \frac{R_0}{R_1} + \frac{R_0}{R_2}\right) \ln \left(16 \frac{R_2}{R_1}\right) = V_G + 3V_T - V_D$$

Resulting Bandgap Voltage

$$V_{Bg} = V_G + 3V_T + V_{os} \left[1 + \frac{R_0}{R_1} + \frac{R_2}{R_3} \left(1 + \frac{R_0}{R_1} + \frac{R_0}{R_2}\right)\right]$$

Nominal Resistor Values

R0 = 160 k	R4 = 385 k	R6 = 50 k
R1 = 320 k	R5 = 325 k	R7 = 800 k
R2 = 320 k	RC = 16 k	R8 = 125 k
R3 = 88 k	RB = 32 k	
	RA = 64 k	